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MPTC for PMSMs of EVs With Multi-Motor Driven System Considering Optimal Energy Allocation

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This paper presents a compound propulsion system with a high-speed permanent-magnet synchronous motor (PMSM) and two in-wheel motors for electric vehicles (EVs). In this study, the longitudinal dynamics model of EVs is firstly presented. Then traction distribution ratio α is introduced to express the traction distribution between the front and the rear axle. Moreover, the function of power consumption in concerned with the traction distribution ratio α is established. Therefore, the α that minimize the power consumption function is selected as the optimal traction distribution ratio. To improve the performance of motor controllers, the model predictive torque control (MPTC) method is employed for high speed and in-wheel motor drives. Experimental comparison with field-oriented control (FOC) shows the advantages of MPTC in dynamic response. Finally, experimental comparisons and HiL tests are presented to verify the MPTC method and the proposed energy allocation method, respectively.

Index Terms—Electric vehicle, permanent magnet synchronous motor, vehicle propulsion system, model predictive torque control

I. INTRODUCTION

With the development of society and economy, environment deterioration and energy shortage are getting more and more concerns. Energy conservation, emission reduction and improving the efficiency of drive system under limited battery capacity have become increasingly important. However, the advantage of the motor drive system cannot be fully utilized as electric vehicles adopts the single power source structure as traditional vehicles [1], [2]. In order to improve the energy efficiency of electric vehicles, propulsion system with multiple traction motors can be adopted. Multi-motor propulsion system can be divided into two types, the four in-wheel motor distributed propulsion system and the front and rear dual motor distributed propulsion system. In [3], R. Wang conducts motor dynamics experiments using four independently driven in-wheel motors and analyzed the energy efficiency characteristics of the wheel motors based on experimental data. Then, an energy distribution method for improving the efficiency of four in-wheel vehicles is proposed. In [4] and [5], N. Mutoh presents the torque distribution strategy of micro-electric vehicles driven by different types of motors in front axle and rear axle. Then the misalignment model of the motor in the constant torque and weak magnetic region are studied, respectively. This strategy can improve the efficiency of motors with different speed range. In [6] and [7], Y. Chen proposes a dual-drive system with speed and torque coupling. Aiming at optimizing the 100-km acceleration time and driving distance, quantum genetic algorithm is used to optimize the parameters of the dynamic system. As each motor in the four in-wheel motors system can be independently controlled, the dynamics

performance of vehicle in this scheme is better than in dual-motor driven system. However, serious safety accidents may be caused, if an in-wheel motor or a motor drive fails at high speed.

In this paper, a composed propulsion system with a high speed PMSM for the front axle and two in-wheel motors for the rear axle is proposed. While, since the in-motor needs to be embedded in the wheel hub, the motor size needs to be reduced. Due to the advantages of high efficiency, high torque density, and widely speed range [8-11], the permanent-magnet synchronous motor (PMSM) is considered as a suitable type of drive motor for in-wheel motors.

The field-oriented control (FOC) method is commonly used for PMSM control. The FOC system usually utilizes cascade control loops with several PI controllers for speed, position and torque control [12], [13]. Due to the advantages of simple algorithm, good robustness and high reliability, the FOC method has been widely employed in PMSM control. However, the cascade structure has its drawback in restricting the overall performance optimization control and disturbance compensation [14-18]. In order to improve the performance of motor drives, the model predictive control (MPC) approach has been developed in recent years. MPC uses mathematic methods to predict the state of the motor at next period according to the motor model, and then the voltage vector of inverter that can minimize the cost function will be selected as a controller output during the next control period [19-25]. There are two types MPC methods usually adopted in motor drives, the model predictive current control (MPCC) method, and the model predictive torque control (MPTC) method. In [25], comparisons between MPTC and MPCC have been carried out and the advantages of MPTC in reducing torque ripples and improving response speed have been pointed out.

In this paper, the MPTC method is employed to control in-wheel motors and the high speed motor. Experimental comparisons with FOC in torque tracking and disturbance rejection conditions have been presented. Moreover,

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hardware-in-loop (HiL) tests are presented to verify the energy efficiency of the proposed propulsion system.

II. VEHICLE LONGITUDINAL DYNAMICS MODEL

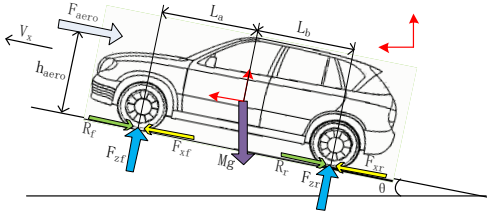


Fig. 1. Vehicle longitudinal dynamics

The torque of motors is mainly related to the load of the vehicle in the forward direction, so the longitudinal dynamics model of the vehicle is mainly concerned. As shown in Fig. 1, the vehicle is moving on a road with the slop of θ at the speed of V_x . The longitudinal dynamics of this vehicle can be described as

$$M \dot{v}_x = F_{xf} + F_{xr} - R_f - R_r - F_{aero} - Mg \sin \theta \quad (1)$$

where F_{xf} , F_{xr} , and F_{aero} are the tire traction force of the front, the rear wheels and the aerodynamic drag force, respectively, R_f and R_r are the rolling resistance of the front and rear wheels, respectively, M is vehicle mass; and g is gravity constant. Some parameters of the vehicle are listed in table I.

Table I

Parameters	Value	Unit
Vehicle mass	900	Kg
Wheelbase	2.2	m
Wheel radius	0.275	m
Transmission ratio	4.75	

The aerodynamic drag force F_{aero} can be expressed as

$$F_{aero} = 0.5 \rho_d C_d A_F (V_x + V_{wind})^2 \quad (2)$$

where ρ_d is the density of air, C_d is the aerodynamic drag coefficient, A_F is the frontal area of the vehicle, and V_{wind} is the wind velocity in the opposite direction of vehicle velocity V_x .

The dynamic equation of a single wheel is expressed as:

$$I_w \dot{\omega}_{f,r} = T_{f,r} - T_b - R_0 F_{xf,xr} - M_R - \zeta \omega_{f,r} \quad (3)$$

where I_w is the tire rotational moment of inertial, $\omega_{f,r}$ is the front or rear tire rotational speed, $T_{f,r}$ and T_b represent traction torque and braking torque of front wheel or rear wheel, respectively, $F_{xf,xr}$ is the longitudinal tire-road friction force, M_R is the moment of rolling resistance, and ζ is the damping coefficient. Since the torque of the front wheel motor is transmitted to the wheels through the transmission system with a ratio of i_0 , the torque of the high speed PMSM is

$$T_f' = \frac{T_f}{i_0} \quad (4)$$

The proposed propulsion system is shown in Fig. 2, and its control system is shown in Fig. 3.

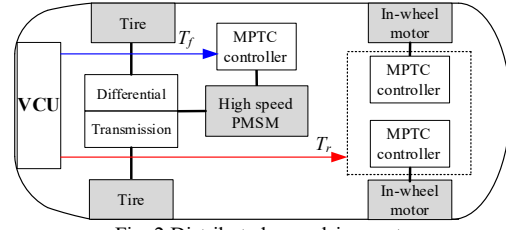


Fig. 2 Distributed propulsion system

III. VEHICLE CONTROLLER

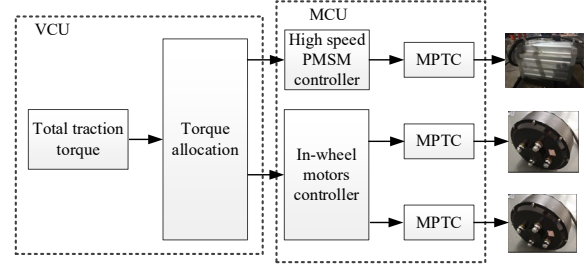


Fig. 3 Block diagram of the propulsion control system

As shown in Fig. 3, The traction force F_x can expressed as

$$\begin{cases} F_x = F_{xf} + F_{xr} \\ F_{xf} = \alpha \cdot F_x = \frac{T_f i_0}{R_0} \\ F_{xr} = (1 - \alpha) \cdot F_x = 2 \frac{T_r}{R_0} \end{cases} \quad (5)$$

where α and $(1-\alpha)$ denote the traction force distribution ratios of the front high speed PMSM and rear in-wheel motors, respectively. Two in-wheel motors are considered under the same traction condition.

The total instantaneous power consumption of three motors can generally be expressed as

$$\begin{aligned} P_{total} &= P_f + 2P_r = \frac{\omega_f \cdot T_f}{\eta_f(\omega_f, T_f)} + \frac{2\omega_r \cdot T_r}{\eta_r(\omega_r, T_r)} \\ &= \frac{F_{xf} \cdot V_x}{\eta_f(V_x, F_{xf})} + \frac{F_{xr} \cdot V_x}{\eta_r(V_x, F_{xr})} \\ &= F_x \cdot V_x \left(\frac{\alpha}{\eta_f} + \frac{1-\alpha}{\eta_r} \right) \\ &= P_{total}(F_x, V_x, \alpha) \end{aligned} \quad (6)$$

where $\eta_f(\omega_f, T_f)$ and $\eta_r(\omega_r, T_r)$ stand for the efficiency of the high speed PMSM and in-wheel motors at their respective operation points, respectively.

It should be noted that, the total power consumption of the propulsion system is determined by total traction force F_x , vehicle speed V_x , and the traction force distribution ratio α . Therefore, the efficiency of the multi-motor driven system can be optimized by selecting an appropriate α value for each operation point (F_x, V_x) , as described in (7)

$$\alpha^* = \arg \min \{ P_{total}(F_x, V_x, \alpha) | 0 \leq \alpha \leq 1 \}$$

s.t.

$$\begin{cases} F_{xf} = \alpha \cdot F_x \leq T_{f,\max} \left(\frac{V_x \cdot i_0}{R_0} \right) \cdot \frac{i_0}{R_0} \\ F_{xr} = (1-\alpha) \cdot F_x \leq T_{r,\max} \left(\frac{V_x}{R_0} \right) \cdot \frac{2}{R_0} \\ 0 \leq V_x \leq V_{\max} \end{cases} \quad (7)$$

The optimal α can thus be obtained by solving the above equation as shown in Fig.4

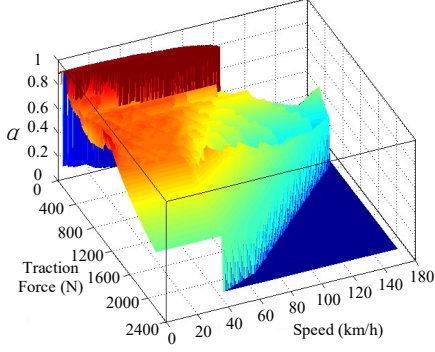


Fig. 4 Optimum traction distribution ratio at different torque and speeds

IV. MOTOR CONTROLLER

A. Dynamic Equations of PMSM

Different from traditional single-motor-driven electric vehicles, the multi-motor-driven electric vehicle can distribute the traction torque more efficiently. Therefore, the motor is required to have a higher response speed. Model predictive control is widely recognized as a high-performance control strategy of PMSM drives due to its quick response and simple principle. While MPTC shows that it is more superior than MPCC at improving system dynamic characteristics.

The model of a surface maintained PMSM in dq -axis rotor frame can be written as

$$\begin{cases} u_d = R_s i_d - \omega_e \psi_q + \frac{d\psi_d}{dt} \\ u_q = R_s i_q + \omega_e \psi_d + \frac{d\psi_q}{dt} \end{cases} \quad (8)$$

the stator flux equation are

$$\begin{cases} \psi_d = L i_d + \psi_f \\ \psi_q = L i_q \end{cases} \quad (9)$$

where $u_d, u_q, i_d, i_q, \psi_d, \psi_q$ are the voltages, currents and magnetic flux of stator in dq -axis, respectively. R_s and L represent the stator resistance and inductance, respectively. ω_e is electromagnetic angular velocity, and ψ_f is the permanent magnet flux linkage.

A discrete-time model of a surface mounted PMSM is described as follow:

$$\begin{cases} \psi_s^{k+1} = T_s u^k + \psi_s^k - T_s i_s^k R_s \\ T_m^{k+1} = \frac{3}{2} P_p \psi_m i_s^{k+1} \end{cases} \quad (10)$$

Thus, the flux linkage and the torque can be calculated directly by obtaining the stator current at time $k+1$. As follows:

$$i_s^{k+1} = i_s^k + \frac{T_s}{L_s} (u_s^k - R_s i_s^k - j\omega \psi_f e^{j\theta}) \quad (11)$$

The estimation value of ψ_s^{k+1} and T_m^{k+1} are utilized for the prediction process. The voltage u_q^k is implemented by the inverter at sampling instant k .

B. Principle of MPTC

While in practical implementation, due to various delay of digital control system, the selected controller output can not be applied to the system until the next instant. This phenomenon will degrade the control performance of the system especially when the sampling frequency is low. Therefore, it is necessary to compensate the impact of one-step delay.

Stator current at time $k+1$ can be calculated from equation (11), and when it is used to calculate torque and flux instead of the current i_s^k , the delay impact will be compensated. However, in order to compensate the one-step delay more effectively, the accuracy of current prediction need to be improved. An improved Euler discretization algorithm is applied

$$\begin{cases} i_p^{k+1} = i_s^k + \frac{T_s}{L_s} (u_s^k - R_s i_s^k - j\omega \psi_f e^{j\theta}) \\ i_s^{k+1} = i_p^{k+1} + \frac{-R_s T_s (i_p^{k+1} - i_s^k)}{2L_s} \end{cases} \quad (12)$$

The cost function is usually concerned with control targets. In EVs, motor controllers are supposed to minimize the flux and electromagnetic torque tracking error.

$$J = |T_m^* - T_m^p| + Q |\psi_s^* - \psi_s^p| \quad (13)$$

where Q is a weighting factor that represents the importance of flux linkage tracking in comparison to torque tracking. For a two-level inverter, there are only eight voltage vectors (including two null vectors). The best voltage vector that minimize the above cost function will be applied in the next control period. The block diagram of proposed MPTC is shown in Fig.5.

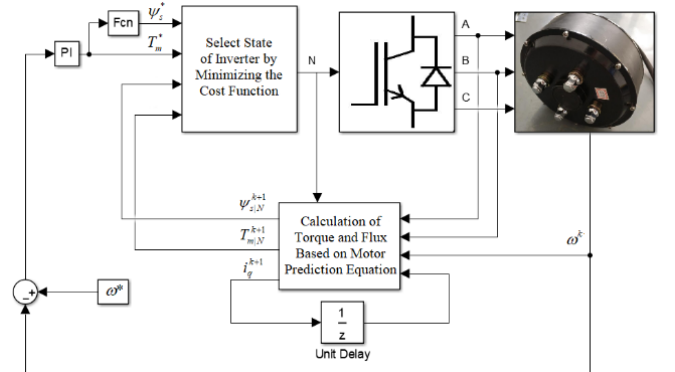


Fig. 5 Block diagram of MPTC

V. EXPERIMENTS

To validate the performance of MPTC controller,

comparison experiments with traditional field oriented control (FOC) method have been carried out. Experimental setups shown in Fig. 6 and Fig. 7 are used to test in-wheel motor controller and high speed PMSM controller, respectively. The test bench consists of a dSPACE DS1401 PPC, a torque sensor, and a magnetic power brake. In order to facilitate the comparison, in-wheel motor controllers and high speed PMSM controller are set to cope with the same traction torque conditions.

First, the vehicle starts accelerating at 0.3s with an acceleration of 2m/s^2 , and the resistance variation are ignored and α is set to 0.6. And the traction torque of high speed PMSM and in-wheel motor can be calculated from (5)

$$\begin{cases} F_x = Ma = 1750\text{N} \\ T_f = \frac{R_0}{i_0} \alpha F_x = 60.8\text{N}\cdot\text{m} \\ T_r = \frac{R_0}{2} (1-\alpha) F_x = 96.25\text{N}\cdot\text{m} \end{cases} \quad (14)$$

Dynamic response of in-wheel motor and high speed PMSM in acceleration condition for MPTC method and FOC method are recorded in Fig. 8.

Next, the motor controllers are tested against sudden external disturbance variations. That is, after the torque of the high speed PMSM and in-wheel motor is stable, a external torque of $10\text{N}\cdot\text{m}$ for 0.1s is applied to them. The torque response under this condition are presented in Fig.9 for MPTC method and FOC method.



Fig. 6 Experimental setup for in-wheel motor



Fig. 7 Experimental setup for high speed PMSM

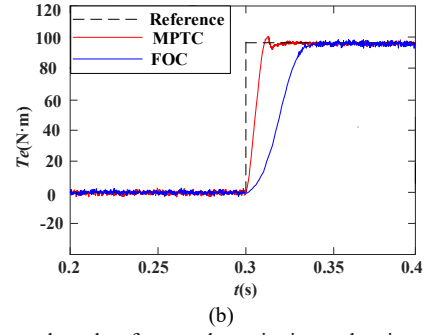
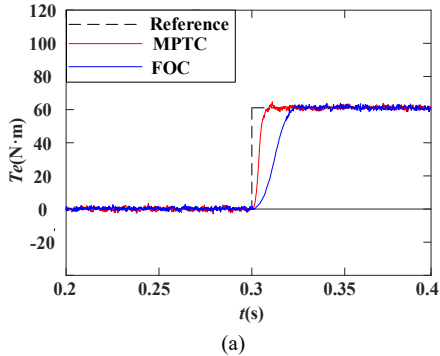


Fig. 8 Experimental results of torque dynamics in acceleration condition (a) In-wheel motor (b) High speed PMSM

Figs. 8 and 9 show the torque response comparisons of the conventional FOC and MPTC under acceleration condition and external disturbance condition, respectively. As can be observed from Fig. 8, when the reference torque is changed in 0.3s, the MPTC in-wheel motor drive takes 0.012s to adjust, while the settling time of FOC in-wheel motor drive is around 0.023s. It can also be noted that, the MPTC high speed PMSM drive can reduce 50% time for adjustment compared to the FOC high speed PMSM drive. As shown in Fig.9, when external disturbance occurs, the MPTC drive can not only reduce the settling time, but also suppress overshoot more effectively. The advantage of MPTC in fast response can improve the acceleration performance of the vehicle directly. Moreover, as α changes according to working conditions, MPTC can help the motor drives to implement the torque distribution strategy more effectively, and the energy consumption can thus be reduced.

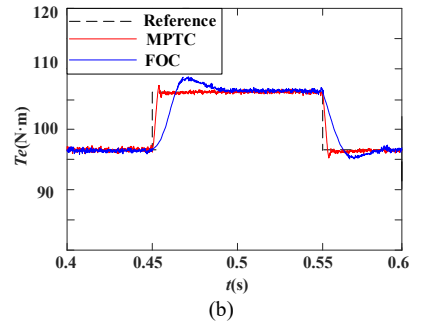
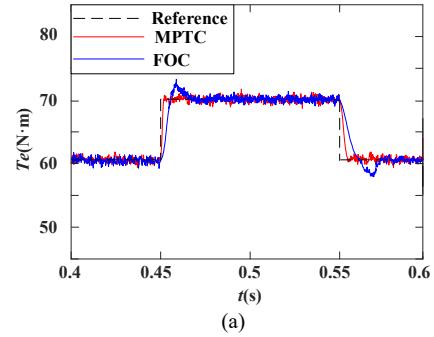


Fig. 9 Experimental results of torque dynamics in external disturbance condition (a) In-wheel motor (b) High speed PMSM

Parameter perturbation and permanent magnet demagnetization are difficult to simulate, thus HiL test is performed to verify the performance of the proposed traction allocation method for the compound EV propulsion system.

HiL test setups are shown in Fig. 10. In addition, the simulink model of the vehicle dynamics and battery management system (BMS) are also combined for energy management.

In order to highlight the superiority of the proposed traction allocation method in energy efficiency, comparison tests in the Urban Dynamometer Driving Schedule (UDDS) with a evenly distributed method have been carried out. With the evenly distributed method, the traction force distribution ratio α is 0.5, i.e., $F_{xf} = F_{xr} = 0.5F_x$.



Fig. 10 HiL setup

Fig. 11 shows the changes of SOC during UDDS. It can be noted that after 5 UDDS, the proposed energy allocation method can save more than 6% energy than the evenly distributed method.

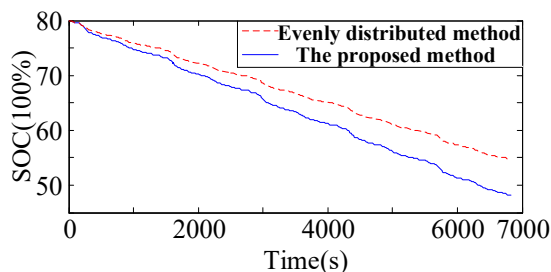


Fig. 11 Changes of SOC

VI. CONCLUSION

This paper presents a MPTC method for a multi-motor driven EV with a high speed PMSM and two in-wheel PMSMs. Moreover, the optimal energy allocation is also considered in this study. The optimal force distribution ratio changes according to the driving condition. Thus, a quick response of motor is required. Experimental tests show that the superiority of MPTC in response speed compared to traditional FOC method. Finally, the HiL tests are carried to verify the advantage of the proposed energy allocation method in energy efficiency compared with the evenly distributed method.

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REFERENCES

- [1] Z. Wang, et al., “Dynamic control of motor torque during the shifting process in PMSM-AMT system.” *IEEE Transportation Electrification Asia-Pacific*, vol. 60, no. 4 pp. 1-5, 2014.
- [2] X. Sun, Y. Shen, S. Wang, G. Lei, Z. Yang, and S. Han, “Core losses analysis of a novel 16/10 segmented rotor switched reluctance BSG motor for HEVs using nonlinear lumped parameter equivalent circuit model”, *IEEE/ASME Trans. on Mechatron.*, vol. 23, no. 2, pp. 747-757, 2011.
- [3] R. Wang, Y. Chen, D. Feng, X. Huang, and J. Wang, “Development and performance characterization of an electric ground vehicle with independently actuated in-wheel motors,” *J. Power Sources*, vol. 196, no. 8, pp. 3962–3971, Apr. 2011.
- [4] N. Mutoh, T. Horigome, and K. Takita, “Driving characteristics of an electric vehicle system with independently driven front and rear wheels,” *IEEE Trans. Ind. Electron.*, vol. 53, no. 3, pp. 803–813, Jun. 2006.
- [5] N. Mutoh and Y. Nakano, “Dynamics of front-and-rear-wheel-independent-drive-type electric vehicles at the time of failure,” *IEEE Trans. Ind. Electron.*, vol. 59, pp. 1488-1499, 2012.
- [6] Y. Chen and J. Wang, “Energy-efficient control allocation with applications on planar motion control of electric ground vehicles,” *Proc. Amer. Control Conf.*, San Francisco, pp. 2719–2724, 2011.
- [7] Y. Chen and J. Wang, “Fast and global optimal energy-efficient control allocation with applications to over-actuated electric ground vehicles” *IEEE Trans. Control Syst. Technol.*, vol. 20, no. 5, pp. 1202–1211, Aug. 2011.
- [8] M. Zeraouia, M. E. H. Benbouzid, and D. Diallo, “Electric motor drive selection issues for HEV propulsion systems: A comparative study,” *IEEE Trans. Veh. Technol.*, vol. 55, no. 6, pp. 1756–1764, Nov. 2006.
- [9] X. Sun, Z. Jin, S. Wang, Z. Yang, K. Li, Y. Fan and L. Chen. Performance improvement of torque and suspension force for a novel five-phase BFSPM machine for flywheel energy storage systems. *IEEE Trans. on Appl. Supercon.*, vol. 29, no. 2, Art. no.: 0601505, 2019.
- [10] Z. Shi, X. Sun, Y. Cai, Z. Yang, G. Lei, Y. Guo, and J. Zhu. Torque analysis and dynamic performance improvement of A PMSM for EVs by skew angle optimization. *IEEE Trans. on Appl. Supercon.*, vol. 29, no. 2, Art. no.: 0600305, 2019.
- [11] X. Sun, Z. Shi, L. Chen, and Z. Yang, “Internal model control for a bearingless permanent magnet synchronous machine based on inverse system method,” *IEEE Trans. on Energy Conversion*, vol. 31, no. 4, pp. 1539-1548, 2016.
- [12] X. Sun, L. Chen, Z. Yang, and H. Zhu, “Speed-sensorless vector control of a bearingless induction motor with artificial neural network inverse speed observer”, *IEEE/ASME Trans. on Mechatronic.*, vol. 18, no. 4, pp. 1357-1366, 2013.
- [13] Y. P. Yang and D. S. Chuang, “Optimal design and control of a wheel motor for electric passenger cars,” *IEEE Trans. Magn.*, vol. 43, no. 1, pp. 51–61, Jan. 2007.
- [14] S. Sue and C. Pan, “Voltage-Constraint-Tracking-Based Field-Weakening Control of IPM Synchronous Motor Drives,” *IEEE Trans. Ind. Electron.*, vol. 55, no. 1, pp. 340-347, Jan. 2008.
- [15] X. Sun, B. Su, S. Wang, Z., G. Lei, J. Zhu, and Y. Guo, “Performance analysis of suspension force and torque in an IBPMSM with V-shape PMs for flywheel batteries,” *IEEE Trans. on Magnetic.*, vol. 54, no. 11, Art. no.: 8105504, 2018.
- [16] X. Sun, L. Chen, H. Jiang, Z. Yang, J. Chen, and W. Zhang. “High-performance control for a bearingless permanent magnet synchronous motor using neural network inverse scheme plus internal model controllers.” *IEEE Trans. on Ind. Electron.*, vol. 63, no. 6, pp. 3479-3488, 2016.
- [17] M. Preindl and S. Bolognani, “Model predictive direct speed control with finite control set of PMSM drive systems,” *IEEE Trans. Power*

- Electron.*, vol. 28, no. 2, pp. 1007-1015, Feb. 2013.
- [18] B. Boazzo and G. Pellegrino, "Model-Based Direct Flux Vector Control of Permanent-Magnet Synchronous Motor Drives," *IEEE Trans. Ind. Appl.*, vol. 51, no. 4, pp. 3126-3136, July-Aug. 2015.
 - [19] M., H., et al., "A Simplified Finite-State Predictive Direct Torque Control for Induction Motor Drive." *IEEE Trans. Ind. Electron.*, vol. 63, no. 6, pp. 3964-3975, 2016.
 - [20] Y. Zhang and W. Xie, "Low Complexity Model Predictive Control—Single Vector-Based Approach." *IEEE Trans. Power Electron.*, vol. 29, no. 10, pp. 5532-5541, Oct. 2014.
 - [21] Y. Zhang and H. Yang, "Generalized Two-Vector-Based Model-Predictive Torque Control of Induction Motor Drives," *IEEE Trans. Power Electron.*, vol. 30, no. 7, pp. 3818-3829, July 2015.
 - [22] Y. Zhang, D. Xu, J. Liu, S. Gao and W. Xu, "Performance Improvement of Model-Predictive Current Control of Permanent Magnet Synchronous Motor Drives," *IEEE Trans. Ind. Appl.*, vol. 53, no. 4, pp. 3683-3695, July-Aug. 2017.
 - [23] Q. Liu and K. Hameyer, "Torque Ripple Minimization for Direct Torque Control of PMSM With Modified FCSMPC," *IEEE Trans. Ind. Appl.*, vol. 52, no. 6, pp. 4855-4864, Nov.-Dec. 2016.
 - [24] N. K. Adamopoulos, , et al., "Comparison of field oriented versus model predictive torque control techniques for monitoring interior PM traction motor over wide speed range," *2017 IEEE 11th International Symposium on Diagnostics for Electrical Machines, Power Electronics and Drives (SDEMPED)*, Tinos, 2017.
 - [25] Geyer, T, "A comparison of control and modulation schemes for medium-voltage drives: emerging predictive control concepts versus PWM-based schemes," *IEEE Trans. Ind. Appl.*, vol. 47, no. 3, pp. 1380–1389, 2011.