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Design of Electrically Small, Huygens Dipole Antenna with Quad-Polarization Diversity

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Abstract — The design of an electrically small, low-profile, Huygens dipole antenna with four reconfigurable polarization states is presented. This very compact design incorporates both electric and magnetic near-field resonant parasitic (NFRP) elements and a reconfigurable driven element. Reconfigurability is achieved with only six PIN diodes. By manipulating their ON/OFF states, this antenna dynamically achieves four polarization states which include two orthogonal linear (LP) and two circular polarization (LHCP and RHCP) states. The simulated values demonstrate that it is electrically small ($ka = 0.944$) and low profile ($0.0449 \lambda_0$) at 1.5 GHz.

Index Terms—Electrically small antennas, low-profile antennas, Huygens dipole antennas, near-field resonant parasitic elements, polarization-reconfigurable antennas.

I. INTRODUCTION

Electrically small antennas (ESAs) with high-directivity have received increasing attention in recent years. They provide significant advantages for many space-limited wireless platforms associated, for example, with long-distance and point-to-point communication systems [1]. To date, various methods for achieving higher directivity in small antenna systems have been widely considered. These include loading the radiating elements with electromagnetic band gap (EBG) structures [2], artificial magnetic conductors (AMC) [3], Quasi-Yagi designs [4-5], Huygens sources [6], and non-Foster elements [7]. The use of NFRP technologies to realize Huygens dipole ESAs has been researched and studied in recent years because these antennas can achieve high directivity and broadside radiation without significantly increasing their size and profile. We have recently developed several electrically small, NFRP, Huygens dipole antennas. These include, for example, linearly polarized and three-dimensional structures [8], multi-layered structures [9], and non-Foster designs [10].

However, it is expected that Huygens source ESAs could be even more useful if they were versatile. For instance, antennas with polarization reconfigurability have many advantages. These include mitigating polarization mismatch, improving system capacity, reducing channel interference, and realizing multiple transmission channels for frequency reuse [11]. Therefore, it would be highly desirable to enable polarization reconfigurability in an electrically small Huygens dipole antenna. Such a polarization diverse ESA would be

suitable for many modern compact wireless applications [12].

A four-state, reconfigurable Huygens source antenna is reported in this paper. The antenna configuration is described in Section II. The simulated results are given in Section III. Finally, some conclusions are drawn in Section IV.

II. RECONFIGURABLE HUYGENS DIPOLE ESA DESIGN

A. Configuration

Fig. 1 illustrates the configuration of the polarization-reconfigurable Huygens source ESA. As shown in Fig. 1, the antenna consists of three substrate layers which are labeled as Layer_1, Layer_2 and Layer_3, respectively. All of these substrates are Taconic TLY-5. All these layers have the same radius, $R1 = 30.0$ mm. However, each of them has a different thickness: $h1 = h2 = 0.25$ mm, and $h3 = 1.58$ mm.

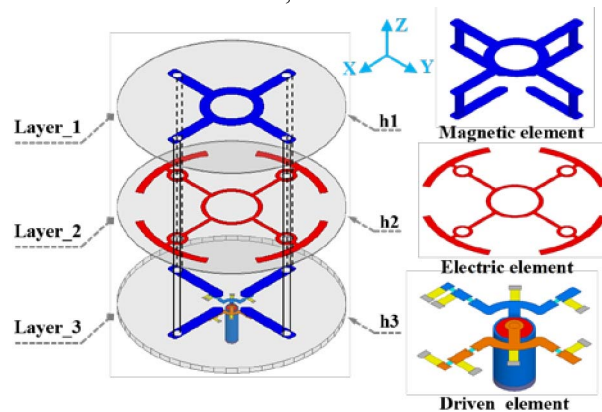


Figure 1. Geometry of the polarization-reconfigurable Huygens dipole ESA.

The magnetic element consists of two intersecting, orthogonally-oriented capacitively loaded loops (CLLs) whose sizes are the same, which has been extensively utilized [13]. One is oriented along the x-axis, and the other is along the y-axis. The upper face of the magnetic element is cross-shaped with a copper ring in the center and is located on the upper surface of Layer_1. Similarly, the electric element consists of two Egyptian axe dipoles (EADs) printed on the upper surface of Layer_2, each having the same size. One is oriented along the x-axis, and the other is along the y-axis. It

also has a copper ring in its center. The coax-driven element lies on the bottom surface of Layer₃.

B. Reconfigurable Driven Element Design

In order to realize the polarization reconfigurability amongst the two LP and two CP modes, a reconfigurable driven element was developed. Its configuration is highlighted in orange and blue in Fig. 2. There are eight additional rectangular copper patches (gray) distributed on the lower surface of Layer₃. As shown in Fig. 2, the Coilcraft inductors are installed between these patches and the driven element strips to block any RF signal from entering into the dc bias network. Each patch is used as the connection point for the dc feeder lines required to control the states of the diodes [14].

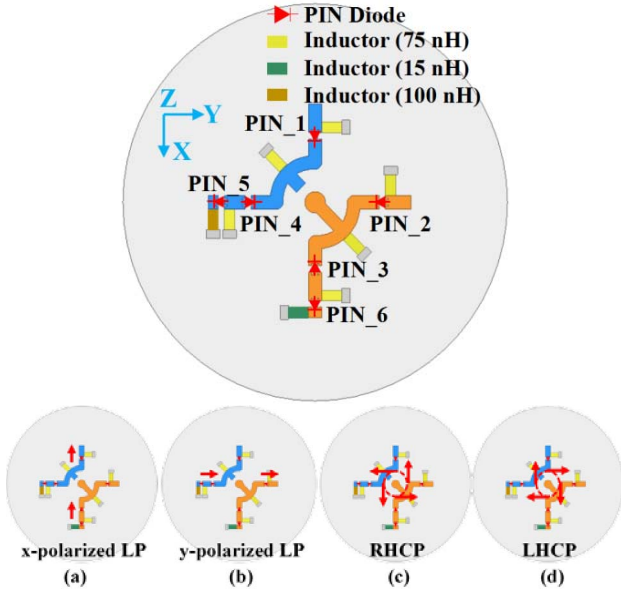


Figure 2. The locations of the six PIN diodes and the associated RF blocking inductors on the driven element. The current directions for each of the four polarization states are indicated by the red arrows in the bottom row of subplots. (a) *x*-polarized LP. (b) *y*-polarized LP. (c) RHCP. (d) LHCP.

Six M/A-COM MA4GP907 PIN diodes were incorporated into this driven element. For an easy description of their ON and OFF states, we have denoted these diodes as PIN₁ - PIN₆ and show their location in Fig. 2. According to the manufacturers datasheet, each diode acts as a 4 ohm resistor in its ON state and as a 0.025 pF capacitor in its OFF state. The control of the radiated polarization with the PIN diode states is summarized in Table I.

TABLE I

THE PIN DIODE STATES TO ATTAIN THE FOUR POLARIZATION STATES

Category	State	PIN ₁	PIN ₂	PIN ₃	PIN ₄	PIN ₅	PIN ₆
Fig.2 (a)	<i>X</i> -LP	ON	OFF	ON	OFF	OFF	OFF
Fig.2 (b)	<i>Y</i> -LP	OFF	ON	OFF	ON	OFF	OFF
Fig.2 (c)	RHCP	ON	ON	ON	ON	ON	OFF
Fig.2 (d)	LHCP	ON	ON	ON	ON	OFF	ON

III. SIMULATED RESULTS OF THE POLARIZATION-RECONFIGURABLE HUYGENS DIPOLE ESA

The simulated S-parameters for the two LP and the two CP polarization states are presented in Fig. 3.

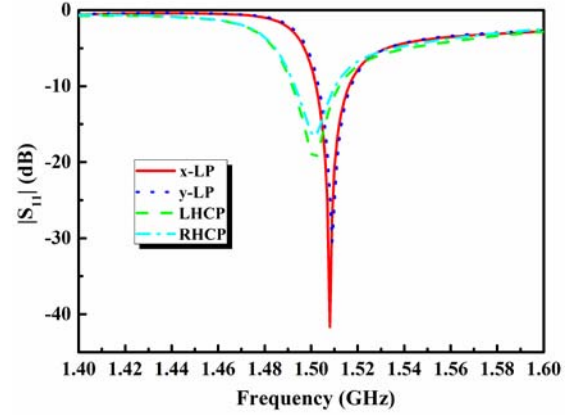


Figure 3. Simulated $|S_{11}|$ values of the four polarization states of the reconfigurable Huygens dipole ESA as functions of the source frequency.

The simulated resonance frequency of the *x*- (*y*-) LP state is 1.508 (1.509) GHz, where $|S_{11}|_{\min} = -41.6$ (-30.0) dB. The simulated -10-dB bandwidths are 1.503-1.516 GHz (13 MHz, 0.86% fractional bandwidth, FBW) and 1.504-1.517 GHz (13 MHz, FBW = 0.86%) for the *x*- and *y*-LP states, respectively. Thus, the simulated overlapped operating bandwidth for the two LP states is 12 MHz, covering 1.504–1.516 GHz (FBW = 0.86%). Similarly, the simulated resonance frequency of the LHCP (RHCP) state is 1.501 (1.501) GHz, where $|S_{11}|_{\min} = -19.5$ (-16.5) dB. The simulated -10-dB bandwidths are 1.492-1.512 GHz (20 MHz, FBW = 1.33%) and 1.493-1.51 GHz (17 MHz, FBW = 1.13%) for these LHCP and RHCP states, respectively. Thus, the simulated overlapped operating bandwidth for the LHCP and RHCP states is 17 MHz, covering 1.493–1.51 GHz (FBW = 1.13%).

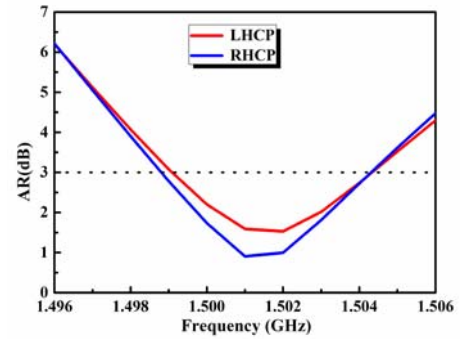


Figure 4. Simulated AR values when the polarization-reconfigurable Huygens dipole ESA is radiating its LHCP and RHCP states.

The axial ratio (AR) values of the two CP states are plotted in Fig. 4 as functions of the source frequency. The simulated 3-dB AR bandwidths are 1.4991-1.5043 GHz (5.2 MHz, FBW = 0.34%) and 1.4988-1.5043 GHz (5.5 MHz, FBW = 0.36%) for the LHCP and RHCP states, respectively. The simulated overlapped operating 3-dB AR bandwidth is 1.4991-1.5043 GHz (5.2 MHz, FBW = 0.34%).

Fig. 5 presents the simulated normalized realized gain patterns when the antenna radiates each of its four polarization states. The main-beam direction for each of these states is clearly broadside, oriented along the $+z$ -axis.

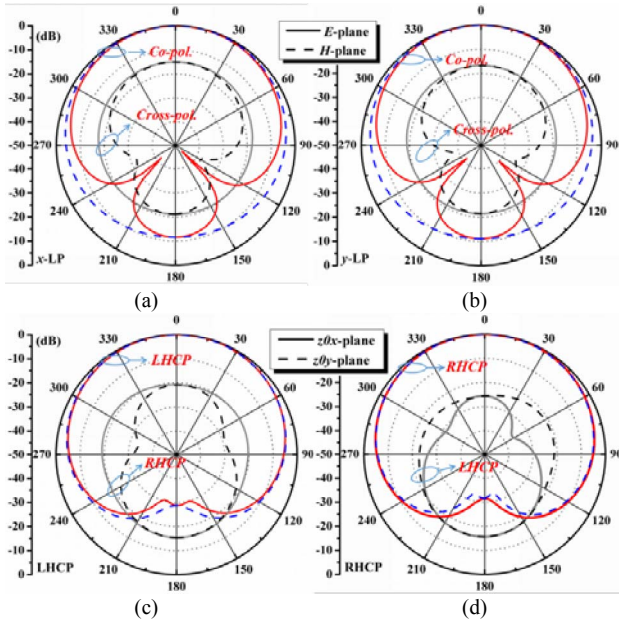


Figure 5. Simulated 2D realized gain patterns for the four polarization states at their resonance frequencies. (a) x -LP. (b) y -LP. (c) LHCP. (d) RHCP.

The simulated normalized realized gain patterns of the x - (y -) LP states are shown in Figs. 5(a) and 5(b), respectively. In the x - (y -) LP state, the simulated peak realized gain, FTBR, and radiation efficiency values are ~ 3.51 (3.54 dBi), ~ 11.2 (11 dB), and ~ 77.9 (78.9%), respectively. The simulated half-power beamwidth is from -61.5° to 62.9° (from -60.5° to 62.3°) in the E-plane and from -76.6° to 76.5° (from -77.4° to 77.7°) in the H-plane. The cross-polarization levels in the broadside direction are below -15 dB for both LP states, indicating their high polarization purity.

The simulated normalized realized gain patterns of the LHCP (RHCP) states are shown in Figs. 5(c) and 5(d), respectively. In the LHCP (RHCP) state, the simulated peak realized gain, FTBR, and radiation efficiency values are ~ 3.13 (2.99 dBi), ~ 15.2 (15.5 dB), and $\sim 71.3\%$ (69.8%), respectively. The simulated half-power beamwidth is from -68.1° to 68.3° (from -68.6° to 70.6°) in the zOx -plane and from -71.1° to 68.7° (from -68.4° to 68.6°) in the zOy -plane.

IV. CONCLUSION

A polarization-reconfigurable Huygens dipole ESA was presented. Rather than altering previously successful Huygens dipole NFRP elements, the reported multi-functional configuration was achieved through the development of a novel, reconfigurable driven element implemented with six PIN diodes. By manipulating their ON/OFF states, it was demonstrated that this ESA is capable of dynamically

achieving four different polarization states, i.e., two LP and two CP states, which are radiated with their maximum gain in the broadside direction. The multi-functionality attained with this polarization-reconfigurable approach could be extended to many other types of reconfigurable antennas. The realized system has a number of potential applications to the next generation wireless systems, e.g., to narrowband MIMO communication systems [15] and narrowband WiFi connections [16].

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