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#### Review 1 Machine Learning for Design Optimization of Electromagnetic 2 **Devices: Recent Developments and Future Directions** 3 Yanbin Li<sup>1</sup>, Gang Lei<sup>2\*</sup>, Gerd Bramerdorfer<sup>3</sup>, Sheng Peng<sup>1</sup>, Xiaodong Sun<sup>4</sup>, and Jianguo Zhu<sup>5</sup> 4 <sup>1</sup> School of Electric and Information Engineering, Zhongyuan University of Technology, Zhengzhou, China; 5 email: liyanbin@zut.edu.cn 6 2 School of Electrical and Data Engineering, University of Technology Sydney, Ultimo, NSW, Australia 2007; 7 8 email: gang.lei@uts.edu.au 3 Department of Electrical Drives and Power Electronics, Johannes Kepler University Linz, Linz, Austria 4040; 9 email: gerd.bramerdorfer@jku.at 10 Automotive Engineering Research Institute, Jiangsu University, Zhenjiang, Jiangsu China 212013; email: 11 xdsun@ujs.edu.cn 12 <sup>5</sup> School of Electrical and Information Engineering, The University of Sydney, Sydney, NSW Australia 2006, 13 email: jianguo.zhu@sydney.edu.au 14 15 \* Correspondence: gang.lei@uts.edu.au; 16 Abstract: This paper reviews the recent developments of design optimization methods for electro-17 magnetic devices, with a focus on machine learning methods. First, the recent advances in multi-18 objective, multidisciplinary, multilevel, topology, fuzzy, and robust design optimization of electro-19

magnetic devices are overviewed. Second, a review is presented to the performance prediction and 20 design optimization of electromagnetic devices based on the machine learning algorithms, including artificial neural network, support vector machine, extreme learning machine, random forest, 22 and deep learning. Last, to meet modern requirements of high manufacturing/production quality 23 and lifetime reliability, several promising topics, including the application of electromagnetic devices. 25

Keywords:electromagnetic devices; electrical machines; optimization methods; machine learning;26deep learning; reliability; topology optimization, robust design.27

## 1. Introduction

Electromagnetic devices have been widely employed in many domestic appliances, 30 biomedical instruments, and industrial equipment and systems, such as electrical drive 31 systems for air conditioners, artificial hearts, electric vehicles (EVs), and more electric air-32 craft, wireless power transmission systems for mobile and EV battery charging, and su-33 perconducting magnetic energy storage (SMES) for power systems. To meet the design 34 specifications and improve their performance, such as high efficiency, high power den-35 sity, and high resource efficiency, optimization is always necessary in the design process. 36 Design optimization of electromagnetic devices has been an active research topic in sev-37 eral international conferences, like COMPUMAG and CEFC. Through extensive research 38 work, many design optimization methods have been employed/developed for electro-39 magnetic devices, including multi-objective, multilevel, and multidisciplinary design op-40 timization methods [1-7]. The performance of electromagnetic devices can be improved 41 by using these methods. 42

As the number of design parameters/objectives and complexity of analysis model 43 increase, high optimization efficiency becomes a serious challenge for many design sce-44 narios, e.g. the multidisciplinary design optimization of machines and drive systems for 45

Citation: Lastname, F.; Lastname, F.; Last-name, F. Title. *Appl. Sci.* 2021, 11, x. https://doi.org/10.3390/xxxxx

Received: date Accepted: date Published: date

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EVs and magnetic levitations (maglevs). The computation costs are huge in many situations due to the high dimension of the optimization problem and the complex multi-physics analysis, e.g., the optimization of a high-speed permanent magnet motor with 10 parameters, 3 objectives, and multi-physics analysis of electromagnetic, thermal and rotor dynamics [8,9]. Therefore, how to improve the optimization efficiency (or reduce the computation cost) is a challenge for efficient design optimization of many electromagnetic devices.

Furthermore, the practical performance of electromagnetic devices is affected signif-53 icantly by the inevitable material diversities and uncertainties in the manufacturing or 54 production process. To improve the manufacturing quality of the optimized electromag-55 netic devices, design optimization in the presence of uncertainties should be conducted at 56 the early stage of the development. From the perspective of industrial production, the 57 performance of a good design of an electromagnetic device, like transformer, should not 58 be sensitive to those uncertainties. To achieve this goal, reliability-based and robust opti-59 mizations have attracted significant research attention recently, especially when the in-60 dustrial big data about the material and manufacturing process are considered [6,10-15]. 61 These topics are of ever-growing significance for smart manufacturing in the context of 62 industry 4.0. However, the application of multidisciplinary analysis and/or industrial big 63 data also brings many challenges to the design optimization process and degrades the 64 optimization performance with conventional optimization methods. Advanced technolo-65 gies, such as machine learning and cloud computing, will greatly improve the handling 66 of these design optimization problems. 67

This paper reviews the recent developments in design optimization of electromag-68 netic devices, with a focus on machine learning methods. Compared to the current state 69 of the art, this review has three new contributions. First, this review covers more types of 70 electromagnetic devices, instead of specific types, like electrical machines and antennas 71 [4,5,713,14]. Second, besides the review of recent developments in typical optimization 72 methods, such as multi-objective and multidisciplinary optimizations, this work reviews 73 the topology optimization, fuzzy optimization, and new optimization strategies like 74 space-reduction strategy. Third, a systematic review of machine learning algorithms is 75 presented and four promising research directions are proposed to integrate these algo-76 rithms with other emerging technologies like digital twin. 77

The remainder of this paper is organized as follows. Section 2 presents an overview 78 of the recent advances in design optimization of electromagnetic devices, including multi-79 objective, multidisciplinary, multilevel, topology, and robust optimization methods. 80 Three examples are investigated, including superconducting magnetic energy storage 81 (SMES), high-frequency transformers, and permanent magnet (PM) motors. Section 3 re-82 views the design optimization of electromagnetic devices based on machine learning 83 methods, with two examples. Section 4 discusses several promising topics as future direc-84 tions for this research field, followed by the conclusion. 85

## 2. An Overview of Recent Advances in Design Optimization of tromagnetic Devices

Design optimization of electromagnetic devices has been an active research topic for many years. Many design optimization methods have been developed through extensive research work worldwide. To compare the performance of different methods, some benchmark works have been developed in International Compumag Society (ICS), like TEAM problems [6,11,12,16-18]. Some papers reviewed popular design optimization methods of several types of electromagnetic devices, such as electrical machines [4-93 6,13,14], and antennas [7]. This section presents an overview of the recent advances in 94 design optimization of electromagnetic devices, including multi-objective, multidiscipli-95 nary, multilevel, topology, and robust optimization methods. Subsection 2.1 starts with 96 the deterministic design optimization (without any consideration of uncertainties). 97

## 2.1. Deterministic Design Optimization

A generic optimization model of the following form can be defined to the multi-ob-99 jective optimization of electromagnetic devices. 100

min: {
$$f_i(\mathbf{x}), i = 1, 2, ..., p$$
}  
s.t.  $g_j(\mathbf{x}) \le 0, j = 1, 2, ..., m$  (1) 101  
 $\mathbf{x}_l \le \mathbf{x} \le \mathbf{x}_u$ 

where *p* and *m* are the numbers of objectives,  $f_i(\mathbf{x})$ , and constraints,  $g_i(\mathbf{x})$ , respectively.  $\mathbf{x}$  is a vector of design parameters,  $\mathbf{x}_i$  and  $\mathbf{x}_u$  are vectors of the lower and upper boundaries of **x**. This model will be simplified as a single-objective problem if *p* is equal to 1. 104

The detailed forms of  $\mathbf{x}$ ,  $f(\mathbf{x})$  and  $g(\mathbf{x})$  depend on the specific type and application of 105 an electromagnetic device. Figure 1 illustrates three popular applications. They are a 106 SMES, a high-frequency transformer, and a surface-mounted permanent magnet synchro-107 nous motor (SPMSM). 108

z(m)line a 10 Inner Outer solenoid solenoid line Inne Solenoid  $h_2/2$ 10 r (m) Outer Solenoid (b) (a) l itz-Wire windings  $D_1$ Magnetic core (c) (d) Stator Core Coil PM Rotor Core  $h_{sy}$  $R_{\rm si}$ (e) (f)

Figure 1. Design optimization illustrations of several electromagnetic devices, (a) a topology of SMES with two 110 solenoids, (b) SMES design structure and optimization parameters, (c) a high-frequency transformer with Litz-111 wire windings, (d) design structure and optimization parameters for the high-frequency transformer, (e) a to-112 pology of an outer-rotor SPMSM, (f) SPMSM design structure and optimization parameters. 113

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SMES is a grid-enabling device for power systems as it can store and discharge a large 114 amount of electricity/power almost instantaneous. SMES stores the power in terms of 115 magnetic energy by its superconducting coils. Nowadays, high penetration of renewable 116 energy sources, like wind and solar, are integrated into the power system worldwide. 117 They will affect the power quality and stability due to their intermittency. This is one of 118 the main challenges of integrating renewable energy sources in the smart grid. SMES is a 119 promising technology to address this challenge [19-21]. The common shapes of supercon-120 ducting coils are solenoid (Figure 1a) or toroidal. The solenoid type is simple, robust and 121 cost-effective. For the design optimization of solenoid-type SMES, there are several pa-122 rameters, such as the dimensions of the solenoids and currents. 123

Figure 1b illustrates an optimization structure of an SMES based on a benchmark 124 problem (TEAM problem 22) in ICS. For this example, eight parameters,  $\mathbf{x} = [R_1, h_1, d_1, J_1, 125, R_2, h_2, d_2, J_2]$ , where (R, h, d) and J are the dimension and current density of the solenoid, 126 respectively, subscript 1 and 2 mean the inner solenoid and outer solenoid, respectively. 127 These parameters will be optimized to minimize the mean stray fields ( $B_{stray}$ ) while keep-128 ing the total stored energy (E) close to 180 MJ. The optimization model can be defined as 129

min: 
$$\begin{cases} f_{1}(\mathbf{x}) = B_{stray} \\ f_{2}(\mathbf{x}) = |E - 180| \\ \\ \text{s.t.} \end{cases} \begin{cases} g_{1}(\mathbf{x}) = |B_{max}| - \min\left(\frac{54 - |J_{i}|}{6.4}\right) \leq 0 \\ \\ g_{2}(\mathbf{x}) = R_{1} + \frac{d_{1}}{2} + \frac{d_{2}}{2} - R_{2} < 0 \\ \\ \\ \mathbf{x}_{l} \leq \mathbf{x} \leq \mathbf{x}_{u} \end{cases}$$
(2) 130

In the model,  $B_{stray}$  is estimated by the magnetic fields on 21 points with the same space 131 along lines *a* and *b*, as shown in Figure 1b. The first constraint is related to the superconductivity of the SMES, where the maximal magnetic field ( $B_{max}$ ) is limited to a value determined by the current density of two coils. This optimization problem can be converted to a single-objective problem (minimizing the mean stray fields only) by considering the requirement of stored energy through a constraint [22-25]. 136

Please note that there are no analytical expressions to show the relationship between 137 design parameters and performance quantities of many electromagnetic devices, for ex-138 ample, the relationship between the parameters x and E in (2). Thus, finite element analy-139 sis (FEA) method is widely employed to calculate the magnetic field distribution. For ex-140ample, Figure 2 shows a design scheme of SMES and its magnetic field distribution by 141 using FEA method (can be done in several software like ANSYS). Due to the symmetry, 142 only the part above x-axis is given. As shown, the maximal magnetic field (indicated as 143 MX) is around 4.27 T. Other performance measures like the energy can be obtained based 144 on the results for the magnetic field. If a parameter, like radius of the inner solenoid, is 145 changed, the corresponding magnetic field and the values of *E*, *B*<sub>max</sub>, and *B*<sub>stray</sub> should vary 146 as well. Thus, FEA and model link the design parameters and performance quantities. 147

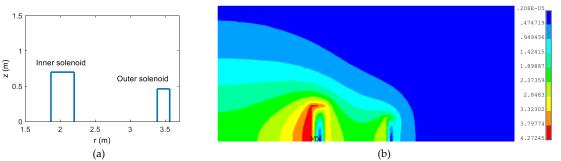


Figure 2. The magnetic field distribution for a design scheme of SMES with current density of 20 MA/m², (a) the149design scheme (symmetric about the x-axis), and (b) the corresponding magnetic field obtained from ANSYS150

Figure 1c shows a prototype of a high-frequency transformer with Litz-wire windings 151 and a magnetic core made of nanocrystalline films. High-frequency transformers have 152 many potential and promising applications, including in the power systems and wireless 153 power transmission systems [26-28]. For the design optimization of a high-frequency 154 transformer, there are many objectives, such as minimizing the loss and volume. Regard-155 ing the design optimization parameters, dimensions (as shown in Figure 1d) and core ma-156 terials (like nanocrystalline or amorphous) can be considered [29-32]. Detailed optimiza-157 tion models can be referred to these works as well. 158

The third example is a PM motor. PM motors have been widely used in industry and 159 transportation, such as hybrid electric vehicles [33-37]. The design optimization of electri-160 cal machines, including MP motors, is very challenging in many situations due to the con-161 sideration of multi-physics analysis. Figure 1e shows the topology of an outer-rotor 162 SPMSM. This kind of machine has been used in many applications as well, like in EVs. In 163 our previous work, it is designed as an in-wheel motor for an EV to achieve four-wheel-164 drive performance [38,39]. This motor has many parameters to optimize, like the dimen-165 sions shown in Figure 1f. In addition, the material of PMs and winding parameters (like 166 the number of turns and winding diameter) can be investigated. Popular optimization 167 objectives are maximizing the output power, average torque, and efficiency, and minimiz-168 ing the cost and torque ripple. 169

Furthermore, as this motor is used as the in-wheel motor, the operating condition 170 should be considered. There are two major challenges for in-wheel-motors, the unsprung 171 weight and cooling [40,41]. The unsprung mass is the weight of all components that are 172 not supported by the suspension, including the wheels with motors, tires, and brakes. As 173 the EV travels up and down over various bumps, potholes, and debris, excessive un-174 sprung weight would cause serious vibration. The weight of in-wheel motors must be 175 minimized (e.g., through topology optimization, will be discussed in section 2.3.5) for 176 smooth drive performance and better vehicle reliability and durability. Cooling is a critical 177 issue for safe operation of high torque density in-wheel motors due to the limited and 178 sealed space in the wheels. Therefore, accurate multi-physics analysis is required, includ-179 ing the electromagnetic, thermal and mechanical analysis. Based on these considerations, 180 an optimization model of this motor can be defined as 181

min:  

$$\begin{cases}
f_1(\mathbf{x}) = -T_{average} \\
f_2(\mathbf{x}) = T_{ripple} \\
f_3(\mathbf{x}) = -\eta \\
f_4(\mathbf{x}) = Mass
\end{cases}$$
(3) 182  
s.t.  

$$\begin{cases}
g_1(\mathbf{x}) = Tem_{pm} - T_0 \le 0 \\
g_2(\mathbf{x}) = Tem_{coil} - T_1 \le 0 \\
g_3(\mathbf{x}) = Vol_m - V_0 \le 0 \\
\mathbf{x}_l \le \mathbf{x} \le \mathbf{x}_u
\end{cases}$$

where  $T_{average}$ ,  $T_{ripple}$ ,  $\eta$ , and Mass represent the average torque, torque ripple, efficiency, 183 and mass of the motor, respectively. The temperature rises in PM ( $Tem_{Pm}$ ), winding 184 (Temcoil), and motor volume (Volm) are considered as constraints. They should not be 185 larger than the limits (indicated as  $T_0$ ,  $T_1$  and  $V_0$ ). For example, for a specific type of PM 186 N38M, its Curie temperature is 100 °C. To avoid demagnetization in operation, T<sub>0</sub> can 187 be defined as 70 °C, assuming that the room temperature in an application is 30 °C. In 188 the implementation of the optimization, both magnetic field analysis and thermal anal-189 ysis should be conducted first to estimate parameters in (3), except the Mass and Volume. 190 Then an optimization algorithm/method can be applied to find the optimal parameters 191 x. Similarly, it is hard to analytically express the relationship between parameters x and 192 many performance quantities in (3), such as torque ripple and efficiency. Therefore, 193 FEA is required for this motor (applies to other motors as well). 194

#### 2.2. Design Optimization Models in The Presence of Uncertainties

Theoretically, the performance of an electromagnetic device can be improved by op-196 timizing the optimization model of (1) or its single-objective form. However, this kind of 197 optimal design (mathematical optimum) often features a lower performance than ex-198 pected after the practical manufacturing process, because there are many inevitable ma-199 terial diversities and uncertainties involved. For example, assume that the optimal height 200 of PMs is 4 mm for an SPMSM after an optimization. Considering a batch production of 201 this motor (for examples, 1,000 motors) with this design scheme, the practical height 202 should be around 4 mm, like 4.05 mm and 3.97 mm, after measurement. It normally fol-203 lows a normal distribution, as indicated by some research work [15]. Therefore, the prac-204 tical performance of this motor will be different from the theoretically optimized value. 205 There are obvious variations in batch production. To improve the manufacturing quality 206 of the motors and other electromagnetic devices, some quality control methods, like six-207 sigma quality control, can be applied. However, this requires a lot of resources which may 208 be a burden for some companies. Alternatively, this problem can be investigated in the 209 early stage of product development through robust design optimization [6,14,42-46]. 210

Figure 3 illustrates a comparison of deterministic and robust optimums, and their per-211 formance variations in the presence of uncertainties. As shown, there are two optima, in-212 dicated as deterministic and robust optimum. For rated conditions, the deterministic one 213 is better than the robust one. However, when a variation  $\Delta x$  occurs, the performance of 214 the deterministic design shows a significant degradation, while some designs likely will 215 not fulfill the illustrated constraint regarding the maximum objective value. This will be 216 regarded as a defect in practical quality evaluation. For example, considering the design 217 optimization of a PM motor, the temperature rise in the winding shall be less than 70 °C. 218 Then, normally, the deterministic design will have an optimum with a temperature rise 219 of the exact 70 °C or very close to it, like, e.g., 69.7 °C. If any uncertainties happen during 220 the manufacturing or operation, the practical temperature rise in the PM may exceed this 221 limit. This may demagnetize the PMs and fail/damage the whole device. By contrast, the 222 robust optimum can ensure the required quality of the device in batch production [6, 47]. 223 That is why the popularity of robust design optimization is increasing compared to the 224 conventional deterministic design optimization in many research fields, including the de-225 sign optimization of electromagnetic devices. 226

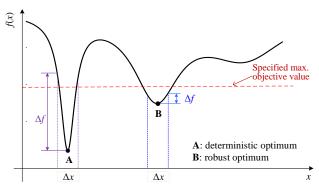


Figure 3. A comparison between the deterministic optimum and robust optimum in the presence of uncertainties for a function f(x), in which  $\Delta x$  and  $\Delta f$  stand for the variations of the parameter and objective, respectively.

There are three popular approaches for the robust design optimization of electromagnetic devices, namely Taguchi parameter design, worst-case design and design for six-233

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sigma (DFSS) [47-54]. Figure 4 shows a block diagram for the Taguchi parameter design 234 method. In this method, the parameters are classified as two groups, control factors and 235 noise factors. Some techniques, such as orthogonal array and signal-to-noise ratios, are 236 then employed to determine the best combination of control factor levels so that the vari-237 ation of this response is minimized in the presence of noise factors [54]. The Taguchi pa-238 rameter design has been widely employed in many applications due to its efficiency and 239 effectiveness. However, there are several drawbacks, e.g., it cannot effectively deal with 240 the constraints in optimization models. 241

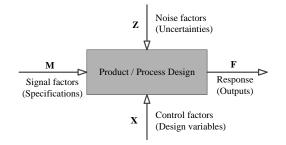


Figure 4. Block diagram of a product/process design in Taguchi method considering control factors (design variables) and noise factors (uncertainties)

Worst-case design and DFSS are able to handle both constraints and optimization 247 objectives in a generic optimization model. Regarding the worst-case approach, its multi-248 objective optimization model can be defined as 249

min: 
$$\{f_{w,i}(\mathbf{x}) = \max_{\xi \in U(\xi)} f_i(\mathbf{x}, \xi), i = 1, 2 ..., p\}$$
  
s.t.  $g_{w,j}(\mathbf{x}) = \max_{\xi \in U(\xi)} g_j(\mathbf{x}, \xi) \le 0, j = 1, 2, ..., m$  (4) 251  
 $U(\xi) = \{\xi \in \mathbb{R}^k | |\xi - \xi_n| \le \Delta \xi\}$ 

where  $\xi$  and  $\xi_n$  are vectors representing the actual and nominal values of noise factors, 253 respectively, and  $U(\xi)$  represents the uncertainty range of these parameters,  $\Delta \xi$  is a vector 254 for the limit of uncertainty range,  $\mathbb{R}$  stands for real coordinate space, k is dimension, sub-255 script *w* in the objective and constraints means the worst case.

For the DFSS approach, its multi-objective optimization model can have the form as 257

min: {
$$F_i[\mu_f(\mathbf{x}), \sigma_f(\mathbf{x})], i = 1, 2, ..., p$$
}  
s.t.  $g_j[\mu_f(\mathbf{x}), \sigma_f(\mathbf{x})] \le 0, j = 1, 2, ..., m$   
 $\mathbf{x}_l + n\boldsymbol{\sigma}_{\mathbf{x}} \le \boldsymbol{\mu}_{\mathbf{x}} \le \mathbf{x}_u - n\boldsymbol{\sigma}_{\mathbf{x}}$   
LSL  $\le \mu_f \pm n\sigma_f \le \text{USL}$ 
(5) 259

where  $\mu$  and  $\sigma$  are the mean and standard deviation, respectively,  $\mu_x$  and  $\sigma_x$  are the mean 260 and standard deviation of x, respectively, LSL and USL are the lower and upper specifi-261 cation limits, respectively. *n* is the sigma level, and it is defined as 6 in many applications. 262 The value of n can be equivalent to a probability of a normal distribution, as shown in 263 Figure 5. Six-sigma level (*n*=6) has been widely adopted in industry as it can provide good 264 reliability for both short-term quality control (equivalent to statistic values) and long-term 265 quality control (with considerations of uncertainties by shifting mean with 1.5  $\sigma$ ). It is 266 equivalent to 0.002 (or a per cent of pass 99.999998%) for short-term quality control and 267 3.4 defects per million opportunities (DPMO) for long-term quality control [6,14,48]. 268

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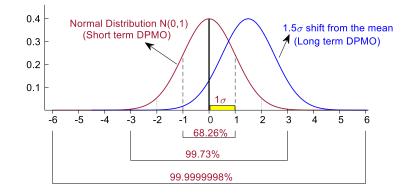


Figure 5. Probability density function of the standard normal distribution for short-term quality control and long-term quality control (with a  $1.5\sigma$  shift from the mean), with probabilities for three sigma levels

As an example, the robust optimization model of the investigated SMES with deterministic optimization model (2) can be defined as 274

min: 
$$\begin{cases} f_{1}(\mathbf{x}) = \mu(B_{stray}) + \sigma(B_{stray}) \\ f_{2}(\mathbf{x}) = \mu(|E - 180|) + \sigma(|E - 180|) \\ \end{cases}$$
s.t. 
$$\begin{cases} g_{1}(\mathbf{x}) = \mu\left(|B_{max}| - \min\left(\frac{54 - |J_{i}|}{6.4}\right)\right) + 6\sigma\left(|B_{max}| - \min\left(\frac{54 - |J_{i}|}{6.4}\right)\right) \le 0 \quad (6) \quad 275 \\ g_{2}(\mathbf{x}) = \mu\left(R_{1} + \frac{d_{1}}{2} + \frac{d_{2}}{2} - R_{2}\right) + 6\sigma\left(R_{1} + \frac{d_{1}}{2} + \frac{d_{2}}{2} - R_{2}\right) < 0 \\ \mathbf{x}_{l} + 6\sigma_{\mathbf{x}} \le \mu_{\mathbf{x}} \le \mathbf{x}_{u} - 6\sigma_{\mathbf{x}} \end{cases}$$

where  $\mu$  and  $\sigma$  are the mean and standard deviation, respectively. In the implementation, 276 each design optimization in **x**, like  $R_1$  (radius of the inner solenoid) can be assumed to 277 follow a normal distribution with two parameters, a mean (the nominal value of a design) 278 and a standard deviation (one third of its manufacturing tolerance) [6,14]. 279

In the implementation of the optimization, the evaluation processes of performance 280 quantities, like E,  $B_{max}$ , and  $B_{stray}$  are same as those applied for solving (2), like FEA. The 281 main difference between (2) and (6) is that some extra information ( $\mu$  and  $\sigma$ ) is needed in 282 (6). To obtain the required data sets, Monte Carlo method can be applied with four main 283 steps. First, assume that each parameter in x follow a normal distribution. Second, gener-284 ate a large amount of samples, like 10,000 samples (means 10,000 design schemes of 285 SMES), from the distributions. Third, evaluate the SMES's performance quantities, such 286 as E and  $B_{max}$ , for these 10,000 designs. Fourth, estimate the mean and standard deviation 287 of these performance quantities. Then optimization algorithms can be applied to find the 288 optimum solutions for this model. 289

There are two main differences between the worst-case approach and DFSS ap-290 proach. First, the worst-case multi-objective model is a minimax optimization problem. It 291 uses the worst motor performance of a design under uncertainties as a measure of robust-292 ness. DFSS uses the sigma level as the measure of robustness. Second, the probability dis-293 tribution functions of the uncertainty parameters are required for DFSS, while the worst-294 case approach only needs intervals for the uncertain parameters. In general, the compu-295 tation cost of worst-case is higher than that of DFSS, as it is a minimax optimization prob-296 lem. Moreover, the worst-case approach is typically more affected by modeling errors, as 297 this quantity is estimated based on a single numerical result, while DFSS measures are 298 determined by evaluating a significant number of design variations. 299

In the case of hybrid uncertainties, the objective functions and constraints have the 300 characteristics of both random and interval uncertainties. Both worst-case and DFSS 301 should be considered in the optimization model, and the computation cost is huge. This 302

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kind of robust optimization has been investigated for a PM motor in our previous work. 303 Polynomial chaos Chebyshev interval (PCCI) method was employed to improve its opti-304 mization efficiency [49]. 305

In the context of Industry 4.0, robust design optimization has been an active and 306 promising research topic in many fields recently, including electrical engineering, me-307 chanical engineering, and civil engineering. A main driving force behind this is that robust 308 design optimization is able to include the manufacturing data and product quality into 309 the design problem. There are many research activities about the robust design analysis 310 and optimization of different types of electromagnetic devices, such as SMES [21,50-53], 311 and several types of electrical machines including high-temperature superconducting lin-312 ear synchronous motor [55], and synchronous reluctance motors [56], and PM motors [57-313 65]. It is observed that there are more discussions on robust design optimization of PM 314 motors than other types of electrical machines, due to the fact that there are many uncer-315 tainties for the PMs. These uncertainties will affect the performance of the PM motors and 316 their reliability, e.g., considering potential demagnetization. This is also challenging for 317 the mass-production of the PM motors. Recently, a special section on robust design and 318 analysis of electric machines and drives was published in the IEEE Transactions on Energy 319 Conversion. Both the robust design analysis and optimization of the motors and control 320 systems are investigated by many authors and, correspondingly, a significant number of 321 papers was published [66]. The outcomes will lay a solid foundation for the development 322 of high-reliability electrical drive systems for many challenging applications, such as EVs 323 and wind power generation. 324

## 2.3. Optimization Methods

## 2.3.1. Optimization Algorithms

After the development of single- and/or multi-objective optimization models, differ-327 ent optimization methods can be employed to discover the optimal results. In general, 328 optimization methods consist of optimization algorithms and strategies. Regarding the 329 optimization algorithms, there are many types, such as gradient-based algorithms and 330 evolutionary optimization algorithms (called intelligent optimization algorithms in many situations). Due to the nature of the high nonlinearity of the optimization models, evolutionary optimization algorithms are more popular nowadays, such as genetic algorithm 333 (GA), differential evolution algorithms (DEAs), PSO algorithm, grey wolf algorithm, ob-334 jective black hole algorithm, and their improvements [39, 67-92]. More details about these 335 optimization algorithms with applications to different electromagnetic devices can be 336 found in review papers [4,5,14]. 337

## 2.3.2. Surrogate Models or Approximation Models

A major challenge for optimizing models (1)-(3) or their single-objectives forms with 339 an appropriate algorithm is the large computation cost, as accurate magnetic field distri-340 bution obtained from 2-D or 3-D finite element analysis (FEA) is required for many appli-341 cations, like PM motors. The FEA usually takes a lot of simulation time, especially for 342 some complex electromagnetic devices that require 3-D finite element models (FEMs). 343 Therefore, surrogate models, such as response surface model (RSM), radial basis function 344 model (RBF), and Kriging model, have been employed to approximate the performance 345 of electromagnetic devices, like flux linkage and core loss. These models can be developed 346 based on the simulation data of FEM by using an appropriate design of experiment (DoE) 347 technique [93-96]. Details about surrogate models and their applications to different elec-348 tromagnetic devices can for instance be found in [4,5,14]. A comparison of different sur-349 rogate models will be discussed in subsection 3.1, with consideration of several machine 350 learning models. Furthermore, these surrogate models can be used to estimate the mean 351

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and standard deviation terms in the robust optimization. This will significantly reduce the computation cost of the typical Monte Carlo analysis with finite element model. 353

## 2.3.3. Multilevel and Space Reduction Optimization Strategies

Surrogate models can be used to improve the optimization efficiency or reduce the 355 computation cost of low-dimensional electromagnetic design problems, for example, in 356 case a total of five dimensions is not exceeded. Its efficiency is not good for high-dimen-357 sional optimization problems, such as the optimization of SMES with 8 parameters (Figure 358 1b) and the optimization of a PM motor with 11 parameters and FEM (Figure 1f). There-359 fore, appropriate optimization strategies should be considered. For this purpose, three 360 optimization strategies, namely multilevel optimization, space reduction optimization, 361 and sequential optimization strategies have been proposed in our previous work for both 362 deterministic/robust and single- or multi-objective optimization problems of electromag-363 netic devices [6,14,47,97-100]. 364

For the multilevel optimization strategy, a high-dimensional optimization problem 365 is converted into several low-dimensional optimization problems by using sensitivity 366 analysis techniques, such as local sensitivity and analysis of variance. Considering the op-367 timization of SMES with 8 parameters, a three-level structure can be defined as: Level 1 (3 368 parameters of [R1, h1, d1]), Level 2 (2 parameters of [J1, J2]), Level 3 (3 parameters of [R2, h2, 369  $d_2$ ]). To implement the optimization, a sequential optimization process, Level 1 – Level 2 370 -Level 3 will be conducted, as shown in Figure 6. This process should be repeated until a 371 convergence criterion is met (for example, the relative error of the objective between two 372 iterations are no more than a given value  $\varepsilon$  like 1%). This kind of optimization strategy will 373 decrease the computation cost, as the optimization of each level is a low-dimensional 374 problem which can be done effectively by a surrogate model. For example, if each factor 375 needs 5 levels in a DoE technique, then Level 1 requires 125 points, Level 2 requires 125 376 points, and Level 3 requires 25 FEM points, resulting in a total of 275 points for one loop 377 of the optimization. If three optimization loops are needed, a total of 825 points are re-378 quired for multilevel optimization. This is much smaller than the samples 379 (125×125×25=390,625) required by developing a model for 8 input parameters. 380

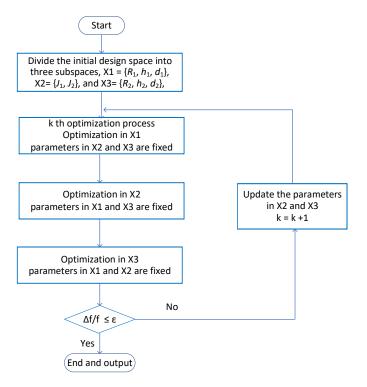


Figure 6. A three-level optimization flowchart for SEMS, in which 8 optimization parameters are allocated to three subspaces X1, X2 and X3.

For the sequential optimization strategy, it uses a sophisticated strategy to sample the most important variants in a small subspace (instead of the initial big design space) by 387 using some space reduction and moving techniques. According to the design examples 388 on SMES and PM motors with soft magnetic composite cores, it can be found that the 389 computation cost of FEA has been reduced significantly by using these strategies. These 390 and improved optimization strategies (like new and improved sensitivity analysis meth-391 ods) have been successfully applied to the design optimization of other PM motors 392 [35,38,101]. 393

## 2.3.4. System-level Multidisciplinary Design Optimization

Besides the optimization problems discussed above, there are two emerging and 395 challenging research topics in the design optimization of electromagnetic devices, system-396 level design optimization and topology optimization. 397

The system-level design optimization is very important for electrical machines and 398 drive systems, e.g., the in-wheel motor drive systems for EVs. The conventional compo-399 nent-level (e.g., the motor) optimization cannot guarantee optimal performance of the 400whole system. To design and optimize this kind of drive systems, electromagnetic analy-401 sis, thermal analysis, mechanical analysis, power electronics, and control systems have to 402 be investigated in the optimization [102-110]. Therefore, multidisciplinary design optimi-403 zation methods should be investigated. Another example is the design of high-speed elec-404 trical machines, where utilizing a multi-physics analysis is crucial to obtain accurate and 405 good optimization results [8,9,111].

## 2.3.5. Topology Optimization

The optimization discussed above is mainly about the structure size or dimension 408 optimization of the electromagnetic devices, which is one of the three main optimizations 409 in engineering, structural size, shape and topology optimizations. The topology optimi-410 zation aims to obtain the optimal layout of components in the design domain for the best 411 objective performance. Compared with the former two optimization methods, the topol-412 ogy optimization is more adept at innovative concept design with superior performance. 413 Moreover, it can shorten the design cycle with less expertise to the optimal design [112]. 414 Topology optimization has been an important research topic in computational electro-415 magnetics for a significant time. It has attracted much attention nowadays due to the re-416 quirements of some modern electromagnetic devices, like the in-wheel motor drive sys-417 tems for (hybrid) EVs, and the development of some advanced AI techniques like deep 418 learning (this will be discussed in the next section) [113-116]. As mentioned in Section 2, 419 unsprung weight is a major challenge for in-wheel-motors [40,41]. The weight of in-wheel 420 motors must be minimized for smooth driving performance and better vehicle reliability 421 and durability. Topology optimization is an effective method to achieve this goal. In many 422 situations, some holes can be designed to the ferromagnetic cores of the motors, such as 423 the rotor cores of IPMSMs (Figure 1e) and the stator cores of SPMSMs (Figure 1g) [112]. 424

There are some challenges for topology optimization as well. To ensure good manu-425 facturability of the obtained design, some constraints, like rounded corners, should be 426 considered in the optimization. Alternatively, this aim can be achieved by robust topology 427 optimization (a combination of robust optimization and topology optimization). 428

## 2.3.6. Fuzzy Optimization

The optimization effectiveness of deterministic and robust models depends on the 430 precise quantifications of design parameters and uncertainties. However, these quantifi-431 cations are not always possible. Fuzzy optimization is good at handling this kind of un-432

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certain. In this case, the performance of electromagnetic devices can be described as qual-433 itative objectives, such as high, medium, and low. Fuzzy membership functions can be 434 used to quantify them and can be included in quantitative optimization models. There are 435 two main types of fuzzy optimization problems in terms of the consideration of con-436 straints. Fuzzy programming has been developed and widely used to handle optimization 437 problems with fuzzy parameters and constraints [117,118]. Fuzzy optimization has been 438 employed to design electromagnetic devices, including different types of motors [119-439 121]. In addition, fuzzy method has been combined with Taguchi method to address 440 multi-objective optimization of electromagnetic devices [122-128]. Taguchi method has a 441 drawback of handling robust multi-objective optimization of electromagnetic devices. 442 Fuzzy method can be employed to solve this kind of problem. 443

## 3. Machine Learning for the Design Optimization of Electromagnetic Devices

From the review and discussions in Section 2, it can be seen that there are two major 445 challenges in the design optimization of electromagnetic devices. First, accurate multi-446 physics analysis is required for many applications, but it normally requires huge compu-447 tation cost of FEA, for example, for the design of high-speed PM motors. Second, highly-448 accurate surrogate models are essential for the optimization process. Naturally, the surro-449 gate models of electromagnetic devices are highly nonlinear. In this case, non-parametric 450 models may be superior to the parametric and semi-parametric models for the perfor-451 mance prediction as there is no specific relationship (like linear) between the inputs and 452 outputs. For example, the relationship between efficiency of a PM motor and its dimen-453 sion may not be able to predict accurately by using polynomials (or RSMs). Fortunately, 454 machine learning presents an opportunity to address these two challenges. 455

Machine learning is a method of data analysis (including prediction and optimiza-456 tion) that automates analytical model building. It is seen as a subset of artificial intelli-457 gence. As a type of non-parametric modeling technique, machine learning is good at de-458 veloping complex nonlinear relationships between a number of inputs and outputs by 459 using different neural networks. Thus, it can be used to build surrogate models for models 460 (1)-(3). Many machine learning algorithms have been used to the design optimization of 461 electromagnetic devices, such as artificial neural networks (ANN), support vector ma-462 chines (SVM), extreme learning machines (ELM), random forest (RF), and deep learning 463 (DL) [3]. DL is a kind of deep neural networks (DNN), and is one subset of machine learn-464 ing algorithms. There are many more layers of neurons in the architectures of DL, com-465 pared with ANN, which can be employed to achieve specialized functionalities. To apply 466 them to design electromagnetic devices, there are two major contributions in the common 467 practice. First, these algorithms have been used to predict/estimate the device's field dis-468 tribution or performance. Second, they can be used to develop surrogate models for opti-469 mization [129-153]. 470

Table 1 lists a comparison of several surrogate models for performance prediction 471 and optimization of electromagnetic devices. There are three types regarding their para-472 metrization. The first category is about parametric models. It includes RSM and RBF. The 473 second one gives semi-parametric models, e.g., Kriging based approaches. The last group 474 involves non-parametric models. It includes three popular machine learning models with 475 explicit mathematical expressions, such as ANN, SVM, and ELM. Please note that RF and 476 DL models are not included in this table as they are hard to be expressed by explicit math-477 ematical equations. 478

These models have been employed to design and optimize different types of electromagnetic devices recently. Please note that different networks may be applied to different machine learning models. For example, there are two popular networks of ANN, backpropagation (BP) and radial basis function networks. And there are many types of DL, such as the convolutional neural network (CNN), recurrent neural network (RNN), and generative adversarial networks (GAN). Table 2 lists some selective bibliography focusing 484 on electromagnetic device design optimization based on different machine learning meth-485 ods. More details are discussed in the following subsections. 486

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## 3.1. Machine Learning for Performance Prediction of Electromagnetic Devices

The performance of electromagnetic devices highly depends on the field analysis re-489 sults of electromagnetic, mechanical, and thermal analyses. These analyses are usually 490 based on FEA and time-consuming, as different dimension and materials and excitations 491 will affect the results. Also, the performance of electromagnetic devices, like the torque 492 and efficiency of a motor, depends on the accurate estimation of the flux linkage and core 493 loss. Typically, those measures feature strongly nonlinear and multi-modal characteristics 494 regarding the input parameters, e.g., due to saturation effects. Consequently, surrogate 495 models based on parametric or semi-parametric approaches might not follow accurate 496 results in general. Through several attempts on machine learning methods, it is found that 497 deep learning algorithms, like CNN and RNN, are good at the distribution estimation of 498 magnetic field and temperature [129-131], and the prediction of torque and efficiency for 499 motors [132-134]. These works established a solid foundation for the generalizable data-500 driven model for the analysis, design and optimization of electromagnetic devices [129]. 501

Model	Mathematical expression	Туре
RSM	$y = X\beta + \varepsilon$ X: structure matrix; $\beta$ : coefficient matrix	Parametric
RBF	$y = \sum_{i=1}^{n} \beta_i H(\ \mathbf{x} - \mathbf{x}_i\ )$ H: RBF function; $\beta_i$ : coefficient matrix	Parametric
Kriging	$y = q(\mathbf{x})'\boldsymbol{\beta} + z(\mathbf{x})$ $q(\mathbf{x})$ : basis function; $\boldsymbol{\beta}$ : coefficient matrix; $z(\mathbf{x})$ : a stochastic process	Semi-parametric
ANN	$y_j = f(\sum_{i=1}^n w_{ji}x_i - \theta_j);$ Basic artificial neuron model, $w_{ji}$ : weightings, $\theta_j$ : neuron's activation threshold; <i>f</i> :transfer function.	Non-parametric
SVM	$y = w \cdot \phi(x) + b$ $\phi$ : a function maps the input space to a higher dimensional feature space, $w$ is a weighting vector, $b$ : bias term.	Non-parametric
ELM	$y = \sum_{i=1}^{K} \beta_i g(\mathbf{w}_i \mathbf{x}_j + b_i)$ g: activation function, <b>w</b> : weighting vector; <b>b</b> : threshold [138].	Non-parametric

Table 1.	Comparison	of several	surrogate models

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An example for the torque prediction of a switched reluctance motor (SRM) based on 505 SVM is considered in the following. Figure 7 shows the machine topology of a four-phase 506 segmented-rotor SRM with 16/10 stator/rotor poles. As shown, the motor consists of 8 507 excited stator poles and 8 auxiliary poles (16 poles in total). The basic operating principle 508 and structural parameters of this motor have been introduced in our previous work [133]. 509 In general, accurate torque modeling of SRM is a difficult problem, as this motor features 510 a double salient structure. Thus, the torque response usually shows a significant ripple, 511 and its modeling is a highly nonlinear problem. In a previous study, two significant fac-512 tors, phase current and position angle, were investigated as inputs for modeling the 513 torque based on three forms of SVM algorithms. They are a conventional SVM, a least 514 square support vector regression (LSSVR) and a maximum-correntropy-criterion-based 515 least squares support vector regression (MCC-LSSVR). Figure 8 illustrates the modeling 516 of phase flux linkage and torque of this motor by using the MCC-LSSVR model. Table 3 517 lists the mean absolute error (MAE) and root mean square error (RMSE) for all three mod-518 eling approaches. As shown, the MCC-LSSVR model appears more effective than the 519 other two techniques. 520

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# **Table 2.** Selective bibliography focusing on electromagnetic device design optimization524based on different machine learning methods525

		base	ed on different machine learning methods
Reference	Model	Application	Estimation objectives or Optimization methods
[129]	CNN	Electromagnetic devices including transformer and PM motor	Magnetic field estimation
[130, 131]	CNN, RNN	PMSMs	Temperature Estimation
[132]	ANN, CNN, RNN	Interior PM motors	Efficiency map and flux-linkage pre- diction
[133, 134]	SVM	Switched reluctance motor	Torque prediction
[135, 136]	SVM	PMSMs	Multi-objective optimization
[137]	RF	Induction machine	Random forest algorithms
[138]	ELM	PM synchronous linear motors	Multi-objective optimization, grey wolf optimization algorithm
[139]	KNN	PM synchronous linear motors	Differential evolution algorithm
[140]	MLP	PMSMs	Hybrid metaheuristic algorithm
[141]	R-DNN	Double secondary linear motor	Cuckoo search algorithm
[142]	CNN	Synchronous reluctance motor	Binary PSO
[143-145]	CNN	Interior PM motors	Topology optimization, multi-objec- tive optimization, genetic algorithm
[146]	ANN	High-frequency transformer	Structure optimization
[147-150]	ANN, SVM, DNN	Antennas	Multi-objective and robust design op- timization

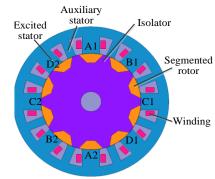


Figure 7. Machine topology of the 16/10 SRM with segmented rotors and excited and auxiliary stators.

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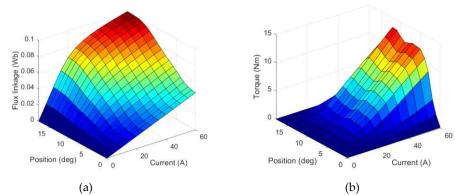


Figure 8. (a) flux linkage prediction based on MCC-LSSVR model with respect to phase current and position angle, (b) torque prediction based on MCC-LSSVR model with respect to phase current and position angle.

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Madalina mathad	Flux linkag	ge (mWb)	Torque (Nm)		
Modeling method	MAE	RMSE	MAE	RMSE	
SVM	0.846	0.756	0.1008	0.0925	
LSSVR	0.424	0.306	0.0525	0.0494	
MCC-LSSVR	0.086	0.073	0.0252	0.0189	

Table 3. Comparison of modeling accuracy of several methods 536

## 3.2. Machine Learning for Optimization of Electromagnetic Devices

Overall, there is more research carried out on machine learning for improving the 538 runtime of the optimization of electromagnetic devices, as was also shown in Table 2. Dif-539 ferent machine learning algorithms, such as SVM, multi-layer perceptron (MLP), KNN, 540 CNN have been investigated to optimize transformers, antennas and motors (motors are 541 the majority applications) [135-142, 146-150]. It is noted that deep learning follows prom-542 ising results when applied for topology optimization of electromagnetic devices and this 543 topic has been attracted much attention recently [143-145]. The presented studies con-544 firmed that good optimization results can be obtained by using different machine learning 545 models for optimization. 546

To illustrate the effectiveness of these models, as an example, a single-objective optimization problem of a SMES is investigated below. Three types of surrogate models, RBF (a parametric model), Kriging (a semi-parametric model), and ANN (a non-parametric 549 model), are compared. Meanwhile, an optimization strategy, sequential optimization 550 method (SOM), is investigated to decrease the computation cost of FEM. 551

In the optimization, the dimensions of the outer superconducting coil,  $[R_2, h_2/2, d_2]$  as 552 shown in Figure 1b, are optimized to minimize the mean stray fields (B<sub>stray</sub>) while keeping 553 the stored energy (E) close to 180 MJ and guaranteeing the requirements for achieving 554 superconductivity. Other parameters are fixed for this case study. Detailed information 555 about the parameters and objective can be found in [98,99]. 556

Figure 9 illustrates the optimization results of SOM by using these three models. As 557 shown, RBF model requires 5 optimization loops to output the final optimum, while the 558 other two models only need 4 loops to converge [6]. Table 4 lists the final optimal. For the 559 purpose of a sound comparison, the direct optimization results of DEA with FEM are 560 listed in the table as well. 561

As shown in the Figure, though the RBF model has the smallest optimum for the first 562 loop of SOM, the differences among the optimal results are small. In the first optimization 563 loop of SOM, the same samples are used to derive the models, then DEA is employed for 564 optimization to find the ideal result. After the convergence of the SOM, the difference 565

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among the considered approaches becomes relatively small, and the overall best results 566 are achieved for the ANN-based approach. Regarding the required number of samples 567 evaluated through FEA, the combination of SOM and any of these modeling approaches 568 necessitates approximately 200 samples, which are less than 10% of that required by the 569 direct DEA optimization (2310 evaluations). Therefore, such approaches are effective for 570 optimization and facilitate minimizing the computational cost and the corresponding 571 runtime. 572

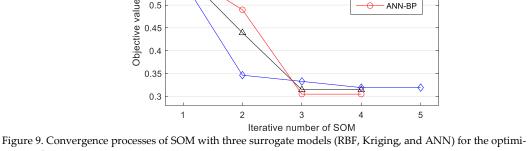
More importantly, there is no big difference between different models with SOM. 573 Therefore, it can be concluded that the optimization strategy may be more important than 574 the particular modeling approach for the optimization of electromagnetic devices. For 575 high-dimensional problems, this conclusion has been confirmed by further studies [47,48]. 576 577

RBF

А

Kriging

ANN-BF



579 zation of SMES 580

I able 4. Optimization results of SMES with different models						
Par.	Unit	DEA	RBF	Kriging	ANN-BP	
R2	m	3.18	3.16	3.11	3.10	
h2/2	m	0.428	0.365	0.267	0.232	
<i>d</i> <sub>2</sub>	m	0.211	0.244	0.340	0.394	
Bstray	mT	1.032	0.957	0.943	0.938	
Ε	MJ	180.00	179.95	179.94	179.94	
F	—	0.344	0.319	0.315	0.313	
FEM		2310	202	157	159	

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## 4. Future Directions

Based on the above discussions, it can be seen that there are many opportunities as 585 well as challenges for the application of machine learning to the design optimization of 586 electromagnetic devices. Compared with conventional design optimization work (includ-587 ing design optimization based on RSM, RBF, and Kriging), the activity of machine-learn-588 ing-based optimization is very limited, as can be seen from Table 2. It is expected that 589 there is a significant increase of corresponding research activities in the future. We think 590 the following topics require further studies: 591

## 4.1. DL for Field Estimation or Multiphysics Analysis

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DL has been successfully employed to estimate the electromagnetic field distribution 593 and temperature distribution of transformers and PM motors, and the efficiency of PM 594 motors. Due to the nature of high nonlinearity, more studies can be conducted to estimate 595 other field distributions for structure analysis. The field estimation of multi-physics anal-596 ysis is challenging for this aspect, especially a coupled field analysis, as for instance re-597 quired for the in-wheel motors and high-speed motors. If multi-physics performance can 598

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be predicted accurately by using DL techniques, this will greatly benefit the optimization 599 work.

## 4.2. Machine Learning for System-Level Design Optimization of Electrical Drive Systems

Machine learning algorithms have been used to optimize the dimensions of several 602 electrical machines, like PM motors. They can be used to design the whole electrical drive 603 systems, including both electrical machines and their power electronics, and control sys-604 tems. Recently, DL has been successfully employed to design the controller to drive the 605 electrical machines [154-155]. As there are many types of control algorithms, such as field-606 oriented control, direct torque control and model predictive control, more research work 607 shall be conducted. 608

Currently, selected machine learning approaches do not show significant advances 609 for solving particular optimization problems involving electromagnetic devices when 610 compared with conventional parametric and semi-parametric modeling techniques. The 611 main reason is that their effectiveness depends on the complexity of the considered opti-612 mization problems, which, for instance, is a function of the number of parameters to be 613 optimized. In case the number of design parameters and objectives and, consequently, the 614 overall complexity of the analysis increases, machine learning algorithms typically feature 615 promising opportunities and feature crucial benefits, e.g., for the system-level multidisci-616 plinary design optimization of electrical drive systems for (hybrid) EVs. 617

## 4.3. Machine Learning for Reliability Improvement of Electromagnetic Devices

High reliability, especially lifetime reliability, is crucial to all electromagnetic devices. 619 Besides the monitoring of devices' operational status, some important work can be done 620 in the stage of design optimization. Many techniques/methods have to be integrated, such 621 as robust topology design optimization, robust tolerance design optimization [156], and 622 multidisciplinary design optimization. Besides the performance modeling, aspects of the 623 manufacturing and the process itself should be considered within the design optimiza-624 tion, like the integrated product and process development of electric drives using a 625 knowledge-based system [157]. 626

Another important technology is the digital twin. Digital twin is an emerging and 627 fast-growing technology which connects the physical and virtual world. It has attracted 628 much attention worldwide recently [158-160]. The future of product and service design 629 will be hugely impacted by digital twin technology. With the help of digital twin, the re-630 liability of the product can be controlled with more freedom. Regarding the design opti-631 mization of electromagnetic devices, it has benefits in three main aspects, product devel-632 opment (design process), manufacturing/production, and operation and management. In 633 the design process, digital twin can be used to test the virtual design scheme given by 634 optimization. Thus, possible design defects can be avoided/corrected in the early design 635 stage of electromagnetic devices. Regarding the production process, digital twin can be 636 applied to determine the best manufacturing process (including product chain and quality 637 control). This will increase the robustness and production efficiency and decrease the pro-638 duction cost of the electromagnetic devices. Regarding the operation and management, 639 digital twin can be employed to find out the best control strategy and parameters for elec-640 tromagnetic devices, like offshore wind generators, to increase their lifetime reliability. 641

## 4.4. Data-Driven Design Optimization Based on Cloud Services

Considering the characteristics and benefits of the technologies mentioned above, a 643 data-driven design optimization platform can be developed based on industrial big data 644 (material data and manufacturing data) and available cloud services (cloud computing 645 [14,161] and manufacturing). In the future, the optimal design of an electromagnetic de-646 vice should include the best topology, shape, dimension and material, and the most ap-647 propriate manufacturing process. Reliability-based design and analysis results should be 648

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available by evaluating a multidisciplinary analysis model and digital twin technology. 649 Machine learning, especially deep learning, will play an important role in this process. 650

## 5. Conclusions

This paper reviewed the recent developments in design optimization of electromag-653 netic devices, with a focus on the application of machine learning algorithms. Through 654 the discussions, it is found that there are many challenges and promising opportunities 655 for the design optimization of electromagnetic devices, with the fast development of ad-656 vanced machine learning algorithms and intelligent manufacturing technology. Besides 657 the requirements of high performance, there are some challenging objectives for the de-658 sign optimization of electromagnetic devices, including high lifetime reliability, high ro-659 bustness and manufacturing quality and flexibility. To address these challenges, there are 660 promising opportunities for the applications of machine learning algorithms and some 661 modern technologies like digital twin. As investigated in Section 4, machine learning al-662 gorithms, such as SVM and DL, revealed very good accuracy in performance prediction 663 of electromagnetic devices, e.g., regarding the estimation of torque and efficiency. DL al-664 gorithms are superior to predict the distribution of the magnetic field and temperature. 665 Due to their excellent suitability for modeling nonlinear characteristics, more extensive 666 research activities on machine learning algorithms are expected in the future. Four prom-667 ising research directions are presented, including the application of cloud services and 668 digital twin, to achieve the intelligent design and manufacturing of electromagnetic de-669 vices with the consideration of lifetime performance and reliability control. 670

Author Contributions: All authors have read and agreed to the published version of the manuscript. 672

Funding: This work was supported in part by the National Natural Science Foundation of China674(NSFC) under grant 61873292, and in part by the Education Department of Henan Province under675project 19A470005.676

Conflicts of Interest: The authors declare no conflict of interest.

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