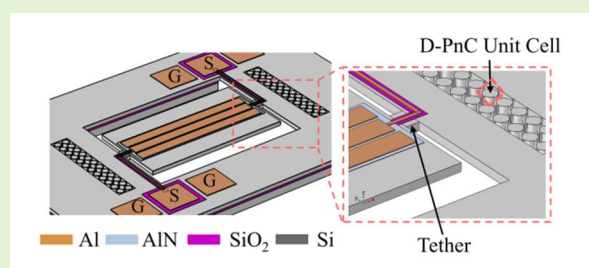


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Effect of Phononic Crystal Orientation on AlN-on-Silicon Lamb Wave Micromechanical Resonators

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Abstract—Phononic crystals (PnCs) have been used to boost the quality factor (Q) of AlN-on-Silicon Lamb Wave Resonators (LWRs). But most reports on applying PnCs to resonators have focused on the common $\langle 110 \rangle$ orientation within (100) silicon. Little is known on the applicability of other crystal orientations. In this work, we explore the effect of orientation on the acoustic band gap (ABG) of two PnC designs and their effect on boosting Q: a disk PnC and a ring PnC. From Finite Element simulation, we show that the disk PnC's ABG is insensitive to orientation while adding a hole into the disk to form a ring changes its ABG to be much more sensitive to orientation. Leveraging the PnCs as anchoring boundary of LWRs, the disk PnC exhibits comparable effectiveness to boost $Q > 11,000$ in the $\langle 110 \rangle$ and $\langle 100 \rangle$ directions while the ring PnC is effective only in the $\langle 110 \rangle$ direction. We further corroborate these trends by incorporating the disk PnC into delay lines in either crystal axis.



Index Terms— Phononic crystal; Orientation; Piezoelectric resonator; Lamb wave; Delay line

I. Introduction

Microelectromechanical Systems (MEMS) resonators have emerged as a promising alternative to quartz crystals in timing applications [1-7]. For oscillator applications, high quality factor (Q) and low motional resistance (R_m) are desired for low phase noise [8-12]. Capacitive resonators can deliver Qs on par with quartz crystals but are limited in their application at higher frequency ranges as high bias voltages are required to reduce an intrinsically large motional resistance (R_m) [13-17]. By contrast, thin-film piezoelectric-on-substrate (TPoS) resonators comprise an AlN film for piezoelectric transduction and a thicker low acoustic loss substrate like silicon [18-23]. TPoS resonators combine the advantages of piezoelectric transduction to deliver low R_m while also having higher Q and power handling from the thick substrate. Besides, TPoS fabrication allows the flexibility to engineer the properties of the substrate to improve the performance of the resonator. For instance, nonlinearity in TPoS resonators can be reduced by tuning the doping concentration and aligning the resonator to the $\langle 100 \rangle$ orientation of the silicon substrate [24]. In addition, it has been shown that the temperature coefficient of frequency

(TCF) of a TPoS resonator can be significantly lowered by fabricating on degenerately doped silicon and aligned to the $\langle 100 \rangle$ orientation [25].

TPoS resonators can have moderately high Qs with careful design to minimize damping, of which anchor loss is rather common to TPoS resonators [26-27]. Among the various approaches to reduce anchor loss in AlN-on-Si piezoelectric resonators [28-34], phononic crystals (PnCs) are an interesting approach for confining acoustic waves within the resonator to boost Q. Most results on the use of PnCs for boosting Q in resonators have been based on the commonly used $\langle 110 \rangle$ axis of (100) Si. Little is known on the application of PnCs in alternative crystal orientations (e.g. $\langle 100 \rangle$) that could be beneficial for other aspects of performance such as lower TCF or higher linearity. Given that Si is anisotropic, changing the orientation of the PnC will generally affect the characteristics of the acoustic band gap (ABG) [35-40] and thus effectiveness in boosting Q.

In this work, we explored the effect of orientation on the acoustic band gap (ABG) of two PnC designs, which will further affect their Q-boosting effectiveness for very high frequency (VHF) band Lamb wave resonator (LWRs) aligned

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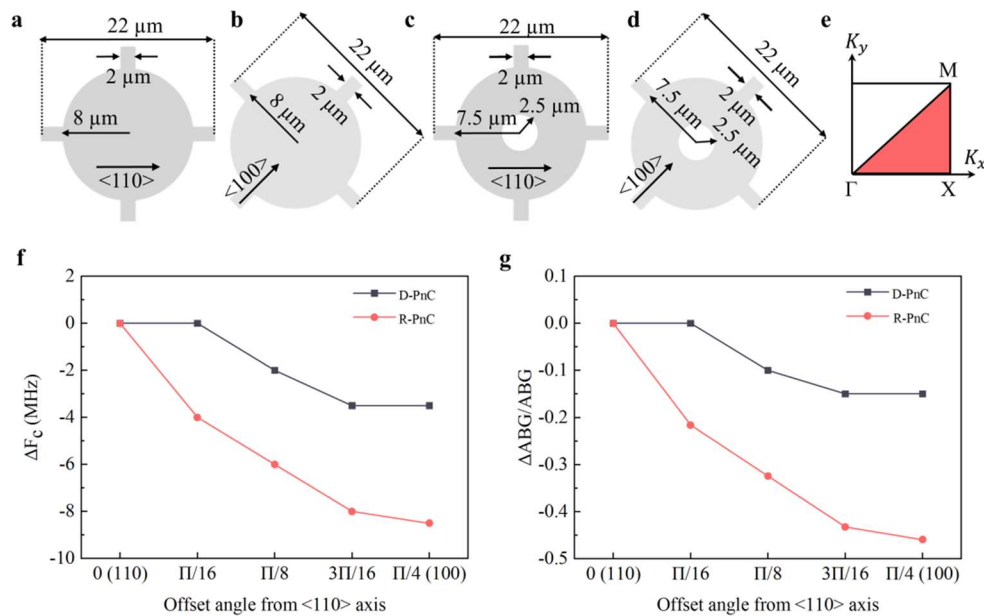


FIG. 1. Top view schematic of a D-PnC unit cell aligned along the (a) $\langle 110 \rangle$ and (b) $\langle 100 \rangle$ directions. Top view schematic of a R-PnC unit cell aligned along the (c) $\langle 110 \rangle$ and (d) $\langle 100 \rangle$ directions. (e) 1st irreducible Brillouin zone in k -space. (f) Change in the center frequency (f_c) of the acoustic bandgap (ABG) as a function of the alignment between the $\langle 110 \rangle$ and $\langle 100 \rangle$ direction within the (100) plane and (g) corresponding fractional change in the size of the bandgap ($\Delta f_{BG}/f_{BG}$) illustrating a larger effect from the R-PnC (45% reduction) over the D-PnC (15% reduction).

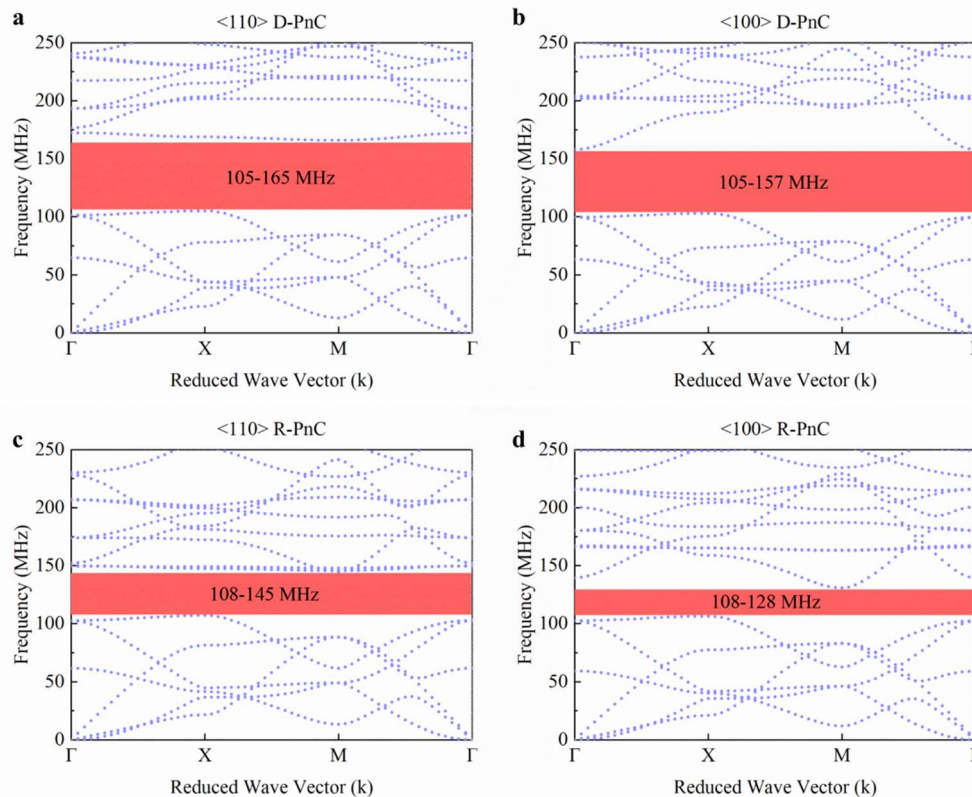


FIG. 2. Simulated frequency band diagrams for the D-PnC aligned along the (a) $\langle 110 \rangle$ and (b) $\langle 100 \rangle$ directions. Simulated frequency band diagrams for the R-PnC aligned along the (c) $\langle 110 \rangle$ and (d) $\langle 100 \rangle$ directions.

to the $\langle 110 \rangle$ and $\langle 100 \rangle$ directions within the (100) plane. The PnC topologies are based on a disk PnC (D-PnC) and a ring PnC (R-PnC). Finite element (FE) simulations show that the D-PnC and R-PnC show different sensitivities to the change in orientation. As a unique case, the ABG size of the D-PnC does not change significantly from one orientation to the other. By

contrast, simply adding a hole in the D-PnC to form an R-PnC causes the ABG to significantly shrink when rotating from the $\langle 110 \rangle$ to the $\langle 100 \rangle$ direction. In addition to the simulation, we fabricate the PnCs into LWRs and delay lines to examine the orientation effect on ABG to boost Q. We show that D-PnCs are equally effective in both the $\langle 110 \rangle$ and $\langle 100 \rangle$ directions

which attain similar levels of recorded high Q [31] while R-PnCs were far less effective in <100> orientation. Moreover, D-PnCs can also provide equally significant attenuation within the ABG when incorporated into delay lines, which further corroborates its excellent effectiveness in Q-boosting for LWRs aligned to either orientation.

II. SIMULATION: EFFECT OF ORIENTATION ON THE ACOUSTIC BANDGAP

The D-PnC and R-PnC unit cell with detailed dimensions described in this work are shown in Fig. 1(a-d). The thickness of both PnCs is defined by a 10 μm silicon layer. We computed the ABG diagram of the PnCs along different crystal orientations within the (100) plane in Si using COMSOL 5.3. For the boundary conditions, we applied Floquet periodic conditions along the x and y directions in each PnC unit domain. Fig. 1(e) illustrates the swept wave vector k bounding the first Brillouin zone. To characterize the effect of orientation on the ABG, we simulated for various orientations between the <110> axis to the <100> axis (45° to the <110> axis) by sweeping the corresponding Euler angles $(\alpha, 0, 0)$ from $(\pi/4, 0, 0)$ to $(0, 0, 0)$ in steps of $(\pi/16, 0, 0)$. The center frequencies (f_c) of the ABGs associated with the D-PnCs and R-PnCs along the <110> orientation were both designed to be around 130 MHz. Fig. 1f illustrates the effect of orientation on the center frequency, where Δf_c denotes the resulting shift in f_c . Fig. 1(g) shows the corresponding effect of orientation on the size of the bandgap (f_{BG}) where Δf_{BG} denotes the resulting change in the size of the bandgap. Rotating from <110> to <100>, D-PnCs see a slight reduction within 15% in f_{BG} , and 3% downshift in f_c . In contrast, the R-PnC sees a far more significant reduction in the f_{BG} by over 45% and a larger downshift by f_c by 7%. We also computed for the GHz and kHz range by scaling the dimensions, and the results are consistent with VHF range. As such, from the simulations, the ABG associated with the D-PnC is rather unique in orientation-insensitivity. Simply adding a hole into D-PnC forming an R-PnC changes the characteristic quite substantially. As shown in [35], the ABG is highly dependent on the constituent materials, especially the contrast between density and elastic constants of the inclusions and the matrix, the geometry of inclusions, the inclusion shape and the fill factor. Despite having the same shape, the D-PnC and R-PnC have two key differences. On one hand, the D-PnC has the matrix of air/silicon and the R-PnC has the matrix of

air/silicon/air. On the other hand, the D-PnC has a larger fill factor than the R-PnC. To some extent, the simpler matrix and larger fill factor may weaken the effect of orientation.

We further simulated the frequency band diagrams of D-PnCs and R-PnCs along the two main orientations: <110> and <100> as depicted in Fig. 2. For D-PnCs shown in Fig. 2(a) and Fig. 2(b), we can observe that it has a wide ABG from 105-165 MHz with f_c around 135 MHz along the <110> direction, and also an equally wide ABG from 105-157 MHz with f_c around 131 MHz along the <100> orientation, which has a slight difference in the f_{BG} width and f_c among these two orientations. In contrast, the change in ABG characteristics is much significant for the R-PnC as shown in Fig. 2(c) and Fig. 2(d). We can see that it has a wide ABG spanning from 108-145 MHz with f_c around 127 MHz along the <110> direction. But the ABG narrows down to 108-128 MHz with f_c shifting down to 118 MHz along the <100> direction, which has an obvious difference in the f_{BG} width and f_c among these two orientations. Thus, we can see the effect of orientation on the ABG of PnCs has a visible difference between different geometries. As expected, orientation in an anisotropic material like Si would notably affect the ABG characteristics as can be seen in the case of the R-PnC. The orientation-insensitivity seen in the D-PnC is likely an exception to the rule.

III. EXPERIMENTAL VERIFICATION: PNCS APPLIED TO LWRs ALIGNED ALONG DIFFERENT ORIENTATIONS

We further fabricated the PnCs into AlN-on-Silicon LWRs as the anchoring boundaries to verify the effect of the PnCs to boost Q along different orientations. The devices in this work were all fabricated by a standard AlN-on-SOI MEMS process [41]. First, we fabricated and tested different number of rows (2 rows, 3 rows and 5 rows) of D-PnC unit cells at the anchoring boundaries of the LWRs along <110> orientation to find an optimal number of rows of unit cells by comparing the resulting Qs. Fig. 3(a) compares the admittance Y_{11} of the three LWRs resonators and Fig. 3(b) are the corresponding Y_{11} spectra after removing feedthrough due to the static capacitance C_0 . We can see that these three LWRs display a similar high level of Q. To this end, we limited the numbers of rows of PnC unit cells to two for a more compact form factor and higher thermal nonlinearity limits owing to lower surface scattering [42]. The unloaded values of Q of all the LWRs herein were derived from

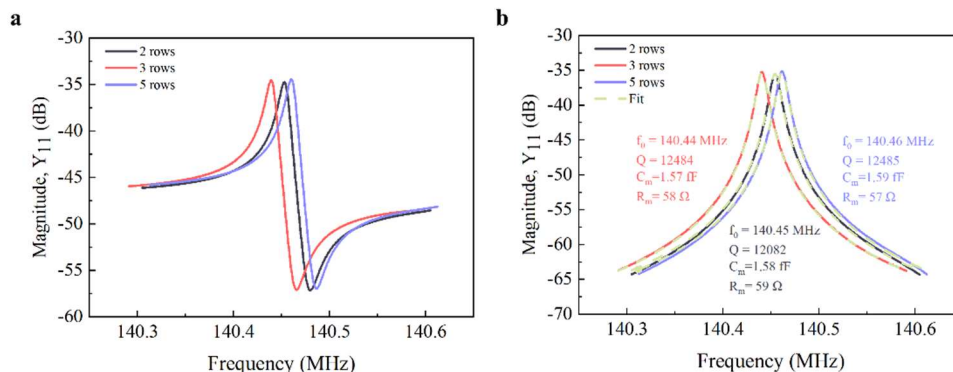


Fig.3. Admittance Y_{11} comparison of the LWRs with different rows of D-PnC as anchoring boundaries (a) before and (b) after removal of feedthrough due to static capacitance C_0 .

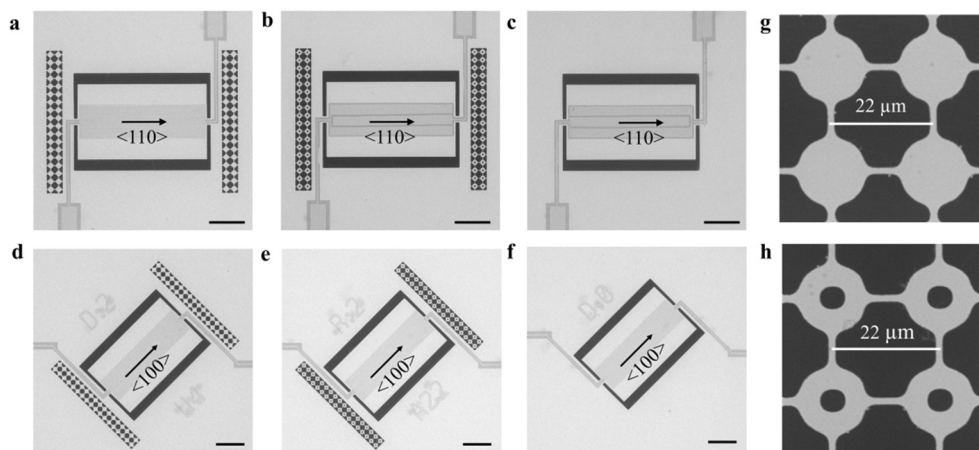


FIG. 4. SEM images of fabricated LWRs aligned to the $\langle 110 \rangle$ direction: a) LWR with D-PnC anchoring boundaries, b) LWRs with R-PnC anchoring boundaries, c) reference LWR with no PnCs at the anchor. (d-f) SEM images of fabricated the same set of three LWRs but aligned to the $\langle 100 \rangle$ direction. Scale bar: 100 μm . SEM images showing a zoom-in view of the g) D-PnC and h) R-PnC with a period or lattice constant of 22 μm .

one-port measurements of LWRs to obtain the S_{11} , from which the admittance Y_{11} was obtained. Q was then extracted by performing a model curve fit based on the modified Butterworth van Dyke model. All the one-port measurements of the LWRs were measured using an Agilent E5061A network analyzer connected through ground-signal-ground probes at ambient pressure and temperature. Open-short calibration was performed prior to all the one-port measurements to determine Q .

To simplify the comparison of the effectiveness in Q -boosting along different orientations, we initially set the D-PnC and the R-PnC to be equally effective by the design of the resonance frequency of LWRs for $\langle 110 \rangle$ orientation. We then characterized the orientation effect by comparing the effectiveness to boost Q of the same LWRs between the same set of PnCs for $\langle 100 \rangle$ orientation. To this end, the orientation is the only variant while the parameters of the PnCs and LWRs keep unchanged. Moreover, the LWRs were designed to be transduced in the 7th order width-extensional mode (140 MHz) for the $\langle 110 \rangle$ orientation, which lies well in the corresponding ABG of the D-PnC (105-165 MHz) and R-PnC (108-145 MHz) for the $\langle 110 \rangle$ orientation. The electrode pitch of the interdigitated transducer (IDT) in all the LWRs is 30 μm . Fig. 4(a)-(c) show three variants of the same LWR designed to align to the $\langle 110 \rangle$ direction. Fig. 4(a) shows an LWR bounded by D-PnCs, Fig. 4(b) shows the same LWR bounded by R-PnCs, and Fig. 4(c) shows the same LWR without PnCs as a reference device for comparison. Similarly, Fig. 4(d) to 4(f) show the same set of three variants but designed to align along the $\langle 100 \rangle$ direction. Fig. 4(g)-(h) depict zoom-in views of the PnCs. The respective resonant frequency ($f_{0,LWR}$) of 7th order width-extensional mode can be calculated from equations (1) and (2):

$$f_{0,LWR} = c/2W_p \quad (1)$$

$$c = \sqrt{E/\rho(1-\nu^2)} \quad (2)$$

where W_p is the pitch size of the electrodes, ρ is the density, E and ν are the Young's modulus and Poisson's ratio respectively. For Silicon, the Young's modulus is 169 GPa for the $\langle 110 \rangle$ orientation and 130 GPa for the $\langle 100 \rangle$ orientation. The Poisson's ratio is 0.064 for the $\langle 110 \rangle$ orientation and 0.28 for the $\langle 100 \rangle$ orientation. The density is 2329 kg/m^3 .

The resonant frequency of LWRs is calculated to be 140 MHz for $\langle 110 \rangle$ orientation and 129 MHz for $\langle 100 \rangle$ orientation respectively, which are consistent with the FE simulations. The LWRs were designed to be only partially covered by IDTs, as seen from Fig. 4(a)-(c) to reduce the effect of electrode-related losses on Q . Fig. 5(a) and Fig. 5(b) compare the admittance Y_{11} of the three LWRs resonators (with D-PnC anchoring boundaries, R-PnC anchoring boundaries, and the reference with no PnC anchoring boundaries) along $\langle 110 \rangle$ and $\langle 100 \rangle$ orientation respectively. Fig. 5(c) and Fig. 5(d) are the corresponding Y_{11} spectra after removing feedthrough due to the static capacitance C_0 . For the LWRs along $\langle 110 \rangle$ orientation illustrated in Fig. 5(c), the LWR with D-PnC anchoring boundaries demonstrates an unloaded Q of over 12000 which is 2.6 times of the reference LWR. The corresponding R_m is reduced by 87 Ω respectively, yielding a 4 dB decrease in insertion loss ($IL=20*\log_{10}[(R_m/2R_0)+1]$, $R_0=50 \Omega$). The LWR with R-PnC anchoring boundaries shows a Q of 11700, which is 2.5 times of the reference LWR. As such, the degree of enhancement from incorporating the D-PnC and R-PnC is similar for $\langle 110 \rangle$ orientations. For LWRs along the $\langle 100 \rangle$ orientation shown in Fig. 5(d), the LWR with D-PnC anchoring boundaries has a Q of over 10000, which is about 2.4 times of the reference LWR. The corresponding R_m is reduced by 68 Ω , yielding a 3 dB decrease in insertion loss. The LWR with R-PnCs anchoring boundaries has a Q of 6890, which is 1.5 times of the reference LWR. Therefore, the R-PnC is less effective than the D-PnC in the Q -enhancement for LWRs along $\langle 100 \rangle$ orientations.

In addition, we also measured another two samples of each

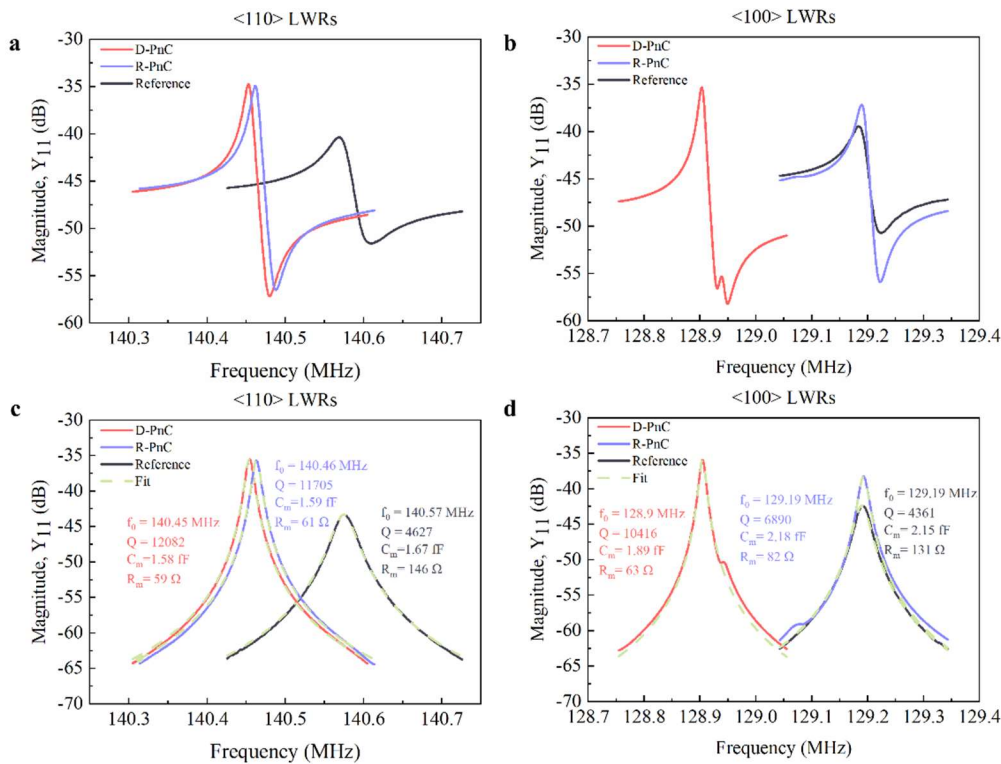


FIG. 5. Y_{11} admittance of each triplet of LWRs (LWR with D-PnC anchoring boundaries, LWR with R-PnC anchoring boundaries and a reference LWR with no PnC at the anchoring boundaries): a) LWRs aligned along the $\langle 110 \rangle$ axis and b) LWRs aligned along with the $\langle 100 \rangle$ axis. Corresponding Y_{11} spectra after removing feedthrough due to the static capacitance C_0 ; c) along the $\langle 110 \rangle$ axis where both PnC-bounded LWRs delivers Q of 12000 (double of the reference LWR), and d) aligned along with the $\langle 100 \rangle$ axis where only the LWR with D-PnC anchoring boundaries delivers a high Q of 10000 that is over double of the reference LWR while the LWR with R-PnC anchoring boundaries delivers a Q of 6900 that is only 1.5 times of the reference LWR.

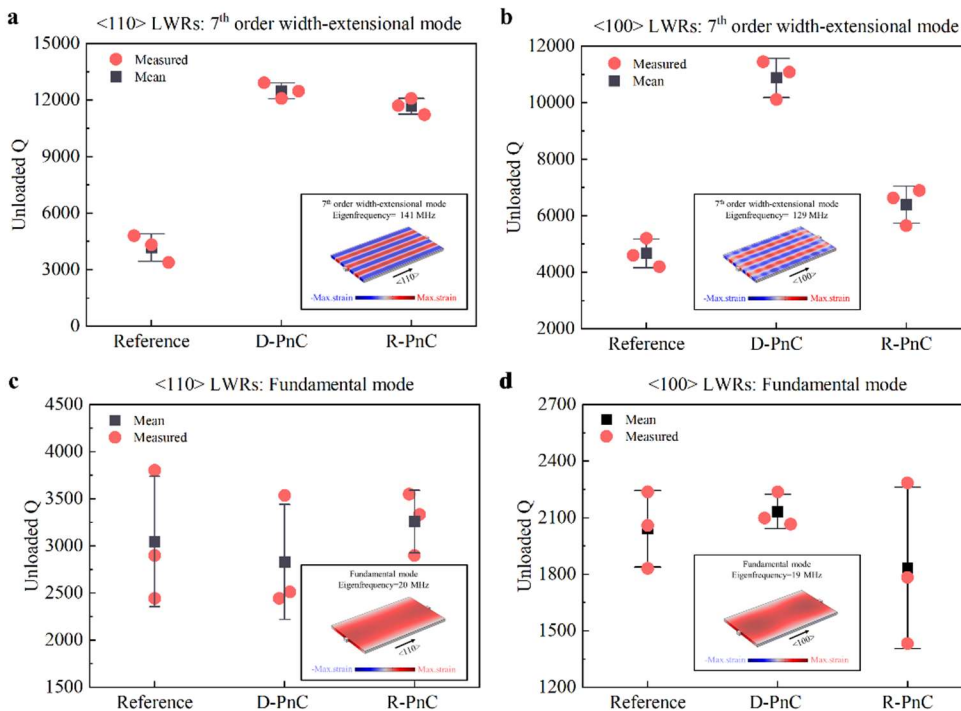


FIG. 6. Comparison of unloaded Q extracted from triplicate samples of each LWR variant in each of the crystal orientations tested transduced in the 7th order width-extensional mode: a) aligned along the $\langle 110 \rangle$ axis where both D-PnC and R-PnC boost Q to over 11000, b) aligned along the $\langle 100 \rangle$ axis where only the D-PnC provides the same significant boost to Q as the $\langle 110 \rangle$ LWRs. Error bars denote the standard deviation. Comparison of unloaded Q extracted from the same three LWRs transduced in the fundamental mode that lies far away from the theoretical ABG: Neither of the PnCs provides a distinct boost to Q for either the c) $\langle 110 \rangle$ axis alignment or d) $\langle 100 \rangle$ axis alignment. Error bars denote the standard deviation.

LWR (i.e. 18 devices in all). Fig. 6(a-b) summarizes the extracted Q from each triplicate of LWRs. For the LWRs with D-PnC anchoring boundaries, the mean Q reaches 12500 along the $\langle 110 \rangle$ orientation and 10900 along the $\langle 100 \rangle$ orientation. Compared to other LWRs fabricated in the same process and of similar frequency, these levels of Q are on par with the highest values we have obtained so far. Where Q reaches the level of over 10000, we have shown that electrode-related losses dominate over anchor loss through extensive modeling [31]. For LWRs with R-PnC anchoring boundaries, the Q was boosted to over 11600 along the $\langle 110 \rangle$ orientation but only increases to 6400 along the $\langle 100 \rangle$ orientation. As such, the D-PnCs are unique in their effectiveness in boosting the Q of LWRs aligned along the main crystal orientations of (100) Si: $\langle 100 \rangle$ and $\langle 110 \rangle$. We did not explore orientations between $\langle 100 \rangle$ and $\langle 110 \rangle$ due to the occurrence of multiple spurious according to FE simulations. Therefore, we have demonstrated that the D-PnCs can provide significant enhancement in Q for the LWRs along either the $\langle 110 \rangle$ or $\langle 100 \rangle$ orientation. In contrast, the R-PnCs can only provide an equivalent level of enhancement for the LWRs along $\langle 110 \rangle$ orientation.

To demonstrate that the PnCs only work inside the ABG, the same LWRs were transduced in the fundamental laterally vibrating mode. For the $\langle 110 \rangle$ direction, this occurs at 20 MHz, while for the $\langle 100 \rangle$ direction, the fundamental mode occurs at 19 MHz, which lie far beyond the ABG of either PnCs. Fig. 6(c-d) summarizes the extracted Q s once again from triplicate measurements of each LWR variant for both crystal axes. As expected from theory, neither the D-PnC nor R-PnC provides any Q -enhancement beyond the ABG along either orientation. As such, the measurement results of the LWRs at the intended 7th order width-extensional mode and at the fundamental mode show that the Q -enhancement by the PnCs is specific to the frequency range of the ABG associated with the PnCs.

IV. EXPERIMENTAL VERIFICATION: ATTENUATION THROUGH PNCS IN DELAY LINES

To compare the effect of orientation on the D-PnC to attenuate the acoustic transmission along each of the $\langle 110 \rangle$ and $\langle 100 \rangle$ axis, we designed and fabricated a pair of delay lines for

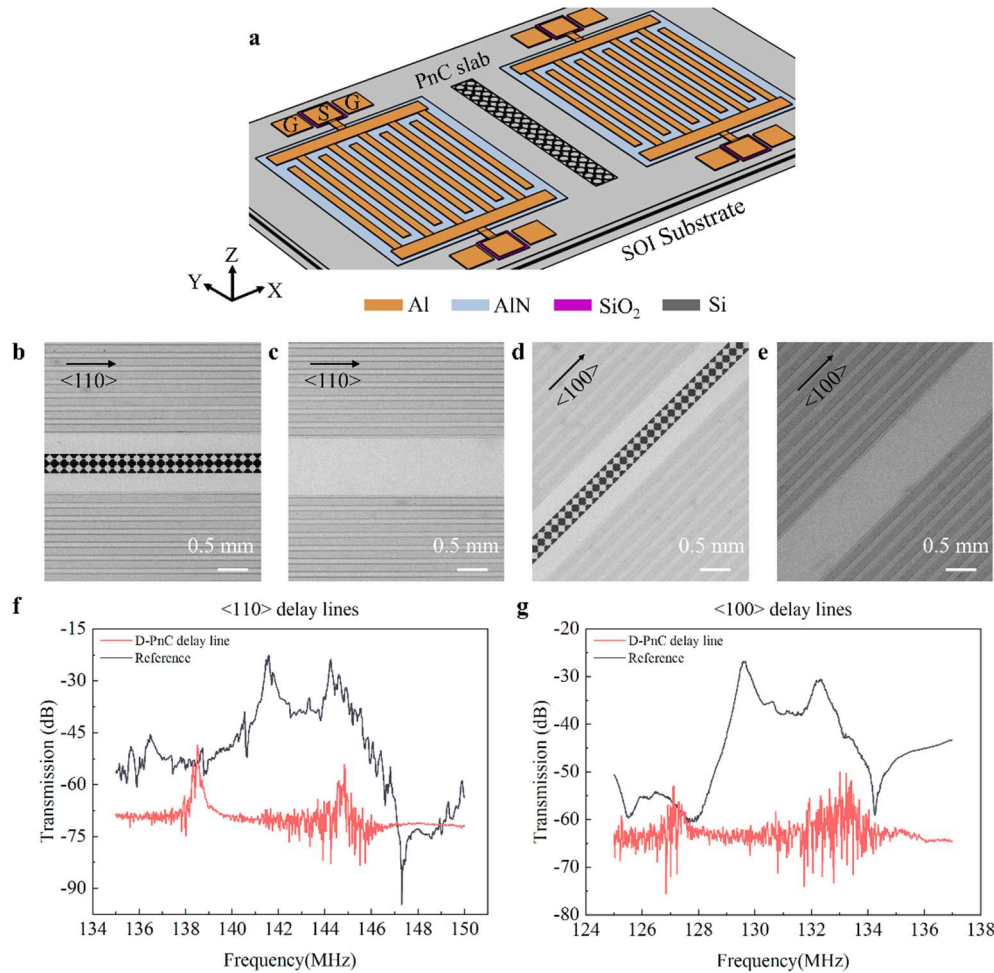


FIG. 7. (a) Schematic of a delay line with PnCs between the transmit and receive IDTs. SEM images of a fabricated delay line along $\langle 110 \rangle$ orientation with (b) D-PnCs in the transmission medium, and a reference delay line with only (c) a solid Si slab in the transmission medium. (d-e) The same delay lines were fabricated for $\langle 100 \rangle$ orientation. Measured electrical transmission (S_{21}) of a delay line incorporating D-PnCs compared against a reference delay line with a Si slab in the transmission medium: (f) delay lines fabricated along the $\langle 110 \rangle$ direction (distinctive transmission band observable between 141-145 MHz) and (g) delay lines fabricated along the $\langle 100 \rangle$ direction (distinctive transmission band from 129-132.5 MHz). In either direction alignments, the D-PnC provides an attenuation >50 dB (full extent masked by noise floor of the instrument).

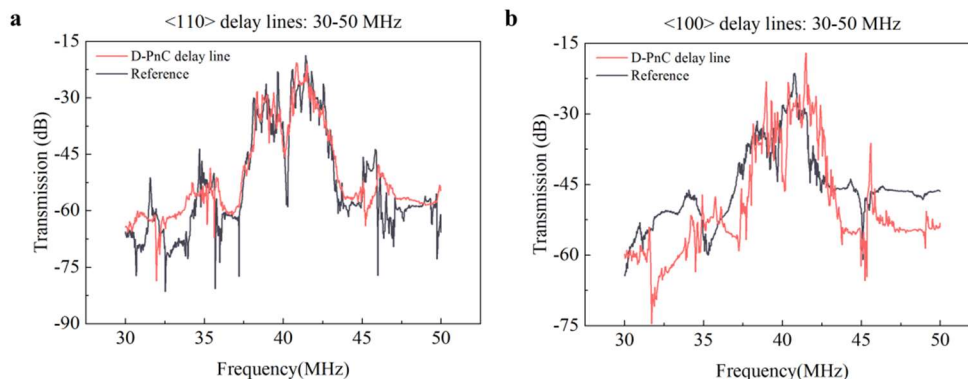


FIG. 8. Comparison of measured transmission (S_{21}) of delay lines with D-PnCs and reference delay lines with a Si slab in the transmission medium when transduced outside the theoretical acoustic band gap in the range of 30-50 MHz. a) Delay lines aligned along the $\langle 110 \rangle$ orientation; b) delay lines aligned along the $\langle 100 \rangle$ orientation. In either orientation alignment, the PnCs do not provide a clear attenuation to wave transmission relative to the reference delay line.

each orientation as shown in Fig. 7(a) (schematic). The pair of delay lines differ only by the defining feature in the transmission medium: D-PnC (Fig. 7(b) and (Fig. 7(d)) and a solid Si slab (Fig. 7(c) and (Fig. 7(e)) as reference delay lines to provide a baseline measurement. As such, a total of four delay lines were designed and fabricated to compare the transmission within and beyond the ABG frequency range of the D-PnC. As the D-PnC has a wide ABG, it is impractical to generate a sufficiently wide transmission through one delay line to cover the full span of the ABG [43]. As such, we utilized the same delay line to generate two limited spans to examine the attenuation inside and outside the ABG. The fabricated delay lines had a pair of standard interdigital differential transducers (IDTs) with a uniform W_p (30 μm as in the case of LWRs). Differential electrodes were employed to reduce parasitic electrical feedthrough by performing fully differential transmission measurements [31]. The IDTs were designed with 18 pairs of fingers with an aperture of 1 mm and the length of the transmission medium was set to 1 mm.

For the span inside the ABG, the delay lines are transduced within the designed center frequency (f_c) along the $\langle 110 \rangle$ and $\langle 100 \rangle$ directions respectively. Fig. 7(f) and Fig. 7(g) compare the measurement results for each pair of delay lines inside the respective ABG along either orientation. We can see that the Si slab delay line shows a distinctive transmission band spanning from 141-145 MHz with a low insertion loss (20 dB) around 142 MHz along the $\langle 110 \rangle$ orientation and a distinctive transmission band spanning from 129-133 MHz with a low insertion loss (26.8 dB) around 129 MHz along the $\langle 100 \rangle$ direction. The down shift in the transmission range corresponds to the difference in the phase velocity in silicon between the $\langle 110 \rangle$ and $\langle 100 \rangle$ directions. In the delay lines incorporating the D-PnCs, we see a huge contrast in the transmission relative to the reference delay lines for either orientation with a similar level of attenuation by 50 dB. These results further corroborate that the D-PnC design is insensitive to the change in orientation and thus exhibits excellent effectiveness in Q-boosting for LWRs aligned to either orientation as described in the previous section.

In addition to the span inside the ABG, the same set of delay

lines were also transduced at much lower frequencies coinciding with the respective fundamental mode, which lies well outside the ABG for either orientation. The experimental results are shown in Fig. 8(a) for the alignment to the $\langle 110 \rangle$ direction and Fig. 8(b) for the alignment to the $\langle 100 \rangle$ direction. In either orientation, we see no visible difference in the transmission curve of the reference delay lines and the delay lines incorporating the D-PnC. As such, these results confirm that outside the ABG, the PnC no longer prohibits the propagation of waves, which corroborates the results of the LWRs transduced in the fundamental mode. Outside the ABG, the PnC no longer confines the waves to the LWR resulting in Q that are not any higher than a reference LWR without PnCs at the anchoring boundaries.

V. CONCLUSION

In this work, we explored the effect of orientation on the acoustic band gap (ABG) of phononic crystals (PnC) to boost Q of AlN-on-Silicon Lamb Wave Resonators. Most reports on PnCs have focused on the commonly used $\langle 110 \rangle$ orientation. Here we studied the effect of changing the orientation of two PnC designs with respect to the ABG and Q-boost of LWRs. We have found that the ABG associated with the D-PnC is rather unique in orientation-insensitivity while simply adding a hole into D-PnC forming an R-PnC changes the ABG to be more sensitive to changes in orientation. Incorporating the D-PnCs into LWRs, we have shown that the D-PnC is equally effective in boosting Q of LWRs to over 11000 for both $\langle 110 \rangle$ and $\langle 100 \rangle$ orientations, a level where electrode-related losses dominate. In contrast, the R-PnC provides the same level of boost to Q only along the $\langle 110 \rangle$ orientation. The equal effectiveness of the D-PnC to boost Q of the LWR in either crystal axis has been further corroborated by measurements of delay lines incorporating these PnCs. The robustness of the D-PnC to different orientation alignments opens up the possibility to engineer resonators along different orientations towards low phase noise and low temperature coefficient of frequency (TCF) resonator-based oscillators.

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